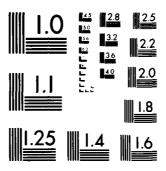
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# NAVAL POSTGRADUATE SCHOOL Monterey, California



### THESIS

A USER'S MANUAL FOR INTERACTIVE LINEAR CONTROL PROGRAMS ON IBM/3033

Ъy

Robert M. Thompson

December 1982

Thesis Advisor:

H. A. Titus

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A User's Manual for Interactive Linear Control Programs on IBM/3033

bу

Robert M. Thompson
Lieutenant, United States Navy
B.S., University of Kentucky, 1975

Submitted in partial fulfillment of the requirements for the degree of

MASTER OF SCIENCE IN ELECTRICAL ENGINEERING

from the

NAVAL POSTGRADUATE SCHOOL December 1982

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### ABSTRACT

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I wish to dedicate this thesis to my wife, Sandi, and to our children, Treva, Melissa, and Robyn, for without their constant love, support, and devotion, this work would not have been possible.

### I. INTRODUCTION

The purpose of this thesis was to update an existing program which provides assistance in solving computational problems associated with the study and application of linear control theory. The Linear Controls Program (LINCON) was first developed by Melsa [ 1] and adapted for batch use at NPS by Desjardins [ 2]. Although the original intent of this thesis was simply to take Desjardins' adapted version of Melsa's LINCON and further adapt it by making it interactive, LINCON soon began to grow as other routines were incorporated, as will be noted, until its present form was achieved.

LINCON, as such, is a high level applications software system made up of a large number of program tools for interactive analysis, design and simulation of a broad class of linear control problems. With LINCON, users can concentrate on their specialized applications rather than system design and routine program development, thereby saving valuable time.

It is assumed that the reader is familiar with the basic concepts of linear control theory as may be obtained from any one of a number of available textbooks (see the

presented in a user-priented fashion. First, their purpose and some general rules that may apply are given; then the input requirements are presented and the expected outputs are described. Examples for each are worked out including a copy of the terminal session and the final results.

### II. SYSTEM DVERVIEW

During LINCON's latest development, underlying guidelines called for concepts which accounted for the needs of the computer system, the programmer, and the user.

The guidelines followed during this latest development stage were:

- (1) Operation of the system should be in an on-line interactive mode such that data can be easily input to (or output from) the system and readily accessed for verification, examination, and processing.
- (2) Program development should be in a high-level language in order to facilitate software implementation and promote machine independence.
- (3) The software should be nodular in structure so that programs can be modified or inserted without affecting existing programs.
- (4) Programs should be invoked by means of logical procedures or commands which minimize interaction time and which are user-oriented so that people can operate the system without first becoming computer experts.

### A. INTERACTIVE OPERATION

The applications of LINCON are focused on interactive processing. Experience has shown that interactive on-line communication has many advantages in a research environment because it offers the opportunity to make observations and select alternate courses of action in a more flexible manner than with batch processing [ 3]. LINCON is organized around a collection of inter-related command programs, each of which performs a specified function and can be executed by means of a simple keyboard initiation sequence.

### B. HIGH LEVEL LANGUAGE

An important feature in the design of LINCON is that it was implemented in a high level language. Program development in assembly language is more time consuming and results in system dependent software.

LINCON is programmed in ANSI standard FORTRAN and follows the conventions of FORTRAN IV. FORTRAN has been found to be a useful language for several reasons:

(1) Since some form of FORTRAN is available on most computers, LINCON is highly portable from one computer to another. A FORTRAN based system is helpful for importing programs as well as exporting

them. Of course, FORTRAN compilers don't all follow the same standards so there can still be difficulties.

- (2) FORTRAN is a simple enough language that relatively complex programs can be implemented in a short period of time. Most scientific and research personnel know FORTRAN sufficiently well to write their own programs if necessary.
- (3) Algorithms can be tested and implemented in FORTRAN and later converted to assembly language versions if more speed and efficiency are necessary. This procedure has the further benefit of aiding portability such that even if parts have been converted to assembler, equivalent FORTRAN versions are available.

#### C. MODULAR SOFTWARE

Overall system flexibilty is achieved by means of modularity. LINCON is actually made up of a large number of independent command programs. Each command program stands on its own with the ability to take some form of input, possibly supplied by some previous command, and generate some form of output, possibly to be used by a follow-up command.

### D. USER-ORIENTED OPERATION

An important aspect in this modification of LINCON was to make the commands user-oriented so that operating the system does not require an engineering or computer background. This was achieved with a standard terminal keyboard by using paonetic characters which relate to the function which the command is to perform. A good combination of brevity and clarity is built into LINCON to avoid having to push extra buttons on the keyboard while at the same time preventing ambiguity.

### E. INPUT RESTRICTIONS AND LIMITATIONS

Although the input requirements are fully described in the presentation of each program, there are several input format similarities used by all of them. For ease of use, and, more honestly, for ease of programming, most of the data input is grouped in the same arrangement.

The first input of every program is used to identify the problem for reference and for output data. A maximum of twenty alpha-numeric characters can be used. This restriction was not a system limitation but a programmer decision.

The next input common to all programs is the dimension of the plant matrix or A matrix. The format is I1 which would normally restrict the user to a maximum matrix size of 9x9, however, due to, again, a programmer decision, a dimension size not to exceed 8 is requested. The reasoning behind this was due, in part, to the printer. The NPS printer is capable of printing 133 characters on a line. Since the output format to the printer is 8E16.6 this naturally limits one to 8 numbers per line. Six places are normally considered necessary for good accuracy. A solution that would have lead to an unlimited matrix size would have been to incorporate a "wrap around" routine within the program. After attempting this it was decided the results were just too difficult to read.

Matrices are entered one element at a time beginning with element 1,1 and continuing across the row. The next row is then entered, and the process continues until all elements have been entered. After the matrix is entered, the complete matrix is automatically brought to the screen for review and possible correction. If a change to the matrix is desired the user simply enters the row number and column number without a separating comma. For example, 35 would

indicate the element in row 3 and column 5. After being prompted the change is entered. A review of the matrix is again screened. The user is again prompted for any possible changes. This procedure continues until all changes have been made.

Any special requirements or limitations will be brought to the user's attention within each program presentation.

## III. MATRIX MANIPILATION, TRANSPER FUNCTION AND TIME RESPONSE PROGRAMS

### A. INTRODUCTION

In this chapter four programs are discussed which may be used for the analysis and design of linear control systems represented in state variable form as

$$\dot{\mathbf{x}}(t) = \underline{\mathbf{A}}\mathbf{x}(t) + \underline{\mathbf{b}}\mathbf{u}(t) \tag{3. A-1}$$

$$u(t) = K[r(t) - k^{T}x(t)]$$
 (3. A-2)

$$y(t) = c^{T}\underline{x}(t)$$
 (3. A-3)

The first, BASMAT, is the Basic Matrix manipulation program which is used for computing the determinant, inverse, characteristic polynomial, and eigenvalues for a square matrix A. In addition, BASMAT will calculate the state transition matrix and the PHI(s) matrix. The second program, PRFEXP, calculates the partial fraction expansion of a polynomial. The third program, ROOFS, calculates the roots of a polynomial. The fourth program is used for determining the time response of linear control systems. RTPESP will determine the rational time response of a system in closed-form provided that the input function r(t) has a

rational time response and that there be no repeated eigenvalues in the combination of the system and input. It should be noted here, and will be again in the actual discussion of the program, that by setting r(t) and K equal to zero, unforced and open-loop systems may be studied, respectfully.

### B. BASIC MATRIX PROGRAM (BASMAI)

Given the plant matrix  $\underline{\mathbf{a}}$ , BASMAT can compute the following:

- (1) the determinant of  $\underline{\mathbf{A}}$ , det  $\underline{\mathbf{A}}$
- (2) the inverse of  $\underline{A}$ ,  $\underline{A}^{-1}$
- (3) the characteristic polynomial, det(sI-A)
- (4) the eigenvalues of the characteristic polynomial,  $\lambda$
- (5) the state transition matrix,  $\frac{1}{2}(t) = \exp(\frac{\Lambda}{L}t)$
- (6) the PHI(s) matrix,  $\underline{\Psi}(s) = (s\underline{\mathbf{I}} \underline{\mathbf{A}})^{-1}$
- 1. Terminal Session Example

This section contains a terminal session for a specific problem. Commands entered by the user are in lower case. All of BASMAT's capabilities will be utilized beginning with following plant matrix:

$$\underline{\mathbf{a}} = \begin{bmatrix} 1.0 & 0.0 & 0.0 \\ 0.0 & -2.0 & 1.0 \\ 0.0 & -0.5 & 1.0 \end{bmatrix}$$

lincon

EXECUTION BEGINS. OF THE FOLLOWING SUBPROGRAMS:

BASIC MATRIX MANIPULATION - <BASMAT>
RATIONAL TIME RESPONSE - <RIRESP>
STATE VARIABLE FEEDBACK - <SIVAR>
CONTROLIABILITY AND OBSERVABILITY - <OBSCON>
LUENBERGER OBSERVER - <LUEN>
OPTIMAL CONTROL/KALMAN FILTERS - <RICATI>
DISCRETE TIME KALMAN FILTER - <KALMAN>
OPTIMAL CONTROL - <OPTION>
PARTIAL FRACTION EXPANSION - <PREEXP>
ROOTS OF A POLYNOMIAL - <ROOTS>
TO USE ONE OF THE SUBPROGRAMS ENTER THE NAME BETWEEN THE

### SYMBOLS < >. basmat

BASMAT PROVIDES MATRIX MANIPULATION TO SOLVE FOR DETERMINANTS, INVERSES, STATE TRANSITION AND PHI(S) MATRICES, EIGENVALUES, AND CHARACTERISTIC POLYNOMIALS.

FIRST ENTER THE PROBLEM IDENTIFICATION (\*NOT TO EXCEED 2) CHARACTERS\*) thesis example

NOW ENTER THE DIMENSION OF THE A MATRIX (UP TO 8).

THE ELEMENT A(1,1) =

THE ELEMENT A(1,2) =

0.0 THE ELEMENT A(1,3) =

0.0 THE ELEMENT A(2,1) =

THE ELEMENT A (2,2) = -2.0

1.0 THE ELEMENT A(2,3) =

0.0 THE ELEMENT A(3,1) =

THE ELEMENT A(3,2) = -0.5

THE ELEMENT A(3,3) =

y

Y

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

DO YOU WANT TO CALCULATE THE DETERMINANT?

THE DETERMINANT OF THE MAIRIX
-1.50E+00

ARE YOU SATISFIED WITH THE RESULTS?

DO YOU WANT THE DETERMINANT PRINTED?

DO YOU WANT TO CALCULATE THE INVERSE?

20

```
THE INVERSE OF THE MATRIX
   ARE YOU SATISFIED WITH THE RESULTS?
y
   DO YOU WANT THE INVERSE PRINTED?
Y
   DO YOU WANT TO CALCULATE THE PHI(S) MATRIX?
    THE MATRIX COEPPICIENTS OF THE NUMERATOR OF THE PHI(5) MATRIX
     THE MATRIX COEFFICIENT OF 5**2
                     0.302+00
     THE MATRIX COEFFICIENT OF 3**1
         1.00E+00 0.0 -2.00E+00 1.00E+00 0.0 -5.00E-01 1.00E+00
     THE MATRIX COEFFICIENT OF 5**0
                     0.3
1.00E+00 -1.30E+00
5.30E-01 -2.00E+00
    ARE YOU SATISFIED WITH THE RESULTS?
A NO RESPONSE WILL GIVE YOU THE OPTION TO MAKE
CHANGES TO THE A MATRIX.
    DO YOU WANT A PRINTOUT OF THE RESULTS?
Y
    DO YOU WANT TO CALCULATE THE CHARACTERISTIC POLYNOMIAL?
     THE CHARACTERISTIC POLYNOMIAL-IN ASCENDING POWERS OF S
          1.50E+00 -2.50E+00 0.0
    ARE YOU SATISFIED WITH THE RESULTS?
   DO YOU WANT THE CHARACTERISTIC POLYNOMIAL PRINTED?
y
    DO YOU WANT TO CALCULATE THE EIGENVALUES?
```

\*\*\*\*\*\*\*\*\*\*

THE EIGENVALUES OF THE A MAIRIX REAL PART IMAG. PART 8.23E-01 0.0 -1.82E+00 0.0 1.00E+00 0.0

ARE YOU SATISFIED WITH THE RESULTS?

DO YOU WANT THE EIGENVALUES PRINTED?

DO YOU WANT TO CALCULATE THE STATE TRANSITION MATRIX?

\*\*\*\*\*\*\*\*\*\*

THE ELEMENTS OF THE STATE TRANSITION MATRIX THE MATRIX COEFFICIENT OF EXP( 8.23E-01)T

THE NATRIX COEFFICIENT OF EXP( -1.82E+00) r

0.0 0.0 1.07E+30 -3.78E-01 0.0 1.39E-01 -6.59E-02

THE MATRIX COEFFICIENT OF EXP ( 1.00E+30) T

ARE YOU SATISFIED WITH THE RESULTS? A NO RESPONSE WILL GIVE THE OPTION TO MAKE CHANGES TO THE A MATRIX.

DO YOU WANT A PRINTOUR OF THE RESULTS?

THIS CONCLUDES THE BASIC MAIRIX MANIPULATION PROGRAM (BASMAT).

ANALYSIS IS COMPLETE. DO YOU WANT TO RUN LINCON AGAIN?
LINCON IS NOW TERMINATED.

The computer results are shown in Appendix A. Interpretation of the determinant, inverse and eigenvalues are straightforward. The PHI(s) matrix is a combination of the numerator coefficients and characteristic polynomial. The results can be interpreted as

$$\mathbf{z}_{(s)} = \frac{1}{s^2-2.5s+1.5} \begin{bmatrix} s^2+s-1.5 & 0.0 & 0.0 \\ 0.0 & s^2-2.0s+1 & s-1 \\ 0.0 & -0.5s+0.5 & s^2+s-2 \end{bmatrix}$$

and, similarily, the fifth term of the state transition matrix is

 $\Phi$ (t) = -0.067exp(3.82t) +1.37exp(-1.82t) -2.26x10-6exp(t)

### C. PARTIAL FRACTION EXPANSION PROGRAM (PRFEXP)

PRFEXP can calculate a partial fraction expansion given a rational ratio of two polynomials in the form

$$G(s) = K - \frac{N(s)}{D(s)}$$
where  $K =$ the input function gain,
$$N(s) = n_1 + n_2 s + n_3 s^2 + \dots + s^3$$
and  $D(s) = d_1 + d_2 s + d_3 s^2 + \dots + s^5$ 

The numerator and denominator coefficients must be arranged so that the coefficients of  $s^{q}$  and  $s^{r}$  are each unity with  $p>q\geq 0$ . The outputs of the program are

- (1) the numerator gain
- (2) the numerator polynomial and its rocts: roots are considered equal if their real and imaginary parts do not differ by more than 0.005.
- (3) the denominator polynomial and its roots; multiple roots are listed once, along with its multiplicity
- (4) the numerator coefficients (residue matrix) are listed in the same order as the denominator mosts; the first coefficient in any row of the matrix is the first-order term coefficient.

### 1. Terminal Session Example

The partial fraction expansion of the following rational polynomial is to be performed:

$$G(s) = \frac{14(s^2+2)(s+1)}{2s^4+5s^3+2s+5}$$

Putting the polynomial in a usable form yields

$$G(s) = \frac{7(s^3+s^2+2s+2)}{s^4+3s^3+s+2.5}$$

lincon

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EXECUTION BEGINS...
LINCON CONSISTS OF THE FOLLOWING SUBPROGRAMS:
BASIC MATRIX MANIPULATION - <BASMAT>
RATIONAL TIME RESPONSE - <RTRESP>
STATE VARIABLE FEEDBACK - <STVAR>
CONTROL LABILITY AND OBSERVABILITY - <OBSCON>
LUENBERGER OBSERVER - <LUEN>
OPTIMAL CONTROL/KALMAN FILTERS - <RICATI>
DISCRETE TIME KALMAN FILTER - <KALMAN>
OPTIMAL CONTROL - <OPTION>
PARTIAL FRACTION EXPANSION - <PRFEXP>
ROOTS OF A POLYNOMIAL - <ROOTS>
TO USE ONE OF THE SJBPROGRAMS ENTER THE NAME BETWEEN THE SYMBOLS <>.
PIFEXP

PRFEXP IS USED TO DETERMINE THE PARTIAL PRACTION EXPANSION OF THE RATIO OF TWO POLYNOMIALS.

FIRST ENTER THE PROBLEM IDENTIFICATION (\*NOT TO EXCEED 23 CHARACTERS\*). thesis example

7. ENTER THE INPUT FUNCTION GAIN--K.

THE INFUT FUNCTION GAIN 7.00E+00

DO YOU WANT TO CHANGE THE VALUE OF THE GAIN?

ENTER THE NUMBERAIDE OF POLYNOMIAL COEFFICIENT OR FACTORED ROOT FORM. FIRST ENTER EITHER THE LETTER PEOR POLYNOMIAL COEFFICIENT FORM OR THE LETTER F. FOR FACTORED ROOT FORM.

25

ENTER THE NUMERALOR POLYNOMIAL ORDER. THE POLYNOMIAL COEFFICIENTS MUST BE ENTERED IN ASCENDING ORDER OF S. \*\*WARNING--THE HIGHEST ORDER COEFFICIENT MUST BE UNITY.\*
DO YOU NEED TO CHANGE THE INPUT FUNCTION GAIN TO SATISFY
THIS REQUIREMENT? n ENTER THE POLYNOMIAL COEFFICIENTS IN ASCENDING ORDER OF S. COEFF(1)= 2. COEFP(2) =2. COEFF(3) =COEFF(4)= 1. NUMERATOR COEFFICIENTS - IN ASCENDING POWERS OF S 2.00E+00 2.00E+00 1.00E+00 NUMERATOR ROOTS ARE REAL PART IMAG. PART 0.0 -1.41E+00 0.0 1.41E+00 -1.60E+00 0.3 ENTER THE DENOMINATOR BY POLYNOMIAL COEFFICIENT OR FACTORED ROOT FORM. FIRST ENTER EITHER THE LETTER P FOR POLYNOMIAL COEFFICIENT FORM OR THE LETTER F FOR FACTORES ROOT FORM. Ď ENTER THE DENOMINATOR POLYNOMIAL ORDER. 4 THE POLYNOMIAL COEFFICIENTS MUST BE ENTERED IN ASCFNDING OPDER OF S. \*\* WARNING -- THE HIGHEST ORDER COEFFICIENT MUST BE INITY. \*
DO YOU NEED TO CHANGE THE INPUT FUNCTION GAIN TO SATISFY
THIS REQUIREMENT? n 

COEFF(2)=

O COEFF(3)=

COEFF(4)=

COEFF(5)=

DENOMINATOR COEFFICIENTS - IN ASCENDING POWERS OF S 2.50E+00 1.00E+00 3.0 3.00E+00 1.00E+00

DENOMINATOR ROOFS ARE

REAL PART IMAG. PART MULTIPLICITY

-3.019E+000.0 1

-9.128E-010.0 1

4.658E-01-8.308E-01 1

4.658E-018.308E-01 1

RESIDUE MATRIX - REAL PART 5.81E+00 3.17E-01 4.36E-01 4.36E-01

RESIDUE MATRIX - IMAG. PART -2.05E-07 0.0 2.06E+00 -2.06E+00

THIS CONCLUDES THE PARTIAL FRACTION EXPANSION PROGRAM (PRFEXP).

DO YOU WANT TO RUN THE PROGRAM AGAIN? n

ANALYSIS IS COMPLETE. DO YOU WANT TO RUN LINCON AGAIN? n

LINCON IS NOW TERMINATED.

The results are shown in Appealix B. Interpretation of these results are:

$$G(s) = \frac{s-5.81+j0}{s+3.02} + \frac{s-0.32}{s+9.13} + \frac{s-0.44-j2.06}{s-0.47+j0.83} + \frac{s-0.44+j2.06}{s-0.47-j0.83}$$

### POLYNOMIAL ROOTS PROGRAM (ROOTS)

ROOTS can calculate the roots of a polynomial of degree less than or equal to eight. Given a polynomial in the form

$$P(s) = p_1 + p_2 s + p_3 s^2 + ... + s^3$$
 (3.D-1)

the coefficient of s\* must be unity. The program output lists the polynomial coefficients for reference and the real and imaginary roots.

### 1. Terminal Session Example

The following polynomial is to be factored:

$$P(s) = -5s^3 + 15s^2 - 25.5s + 4$$

Putting the polynomial in the required form yields

$$P(s) = s^3 - 3s^2 + 5.1s - 0.8$$

lincon

EXECUTION BEGINS...

LINCON CONSISTS OF THE FOLLOWING SUBPROGRAMS:

BASIC MATRIX MANIPULATION - <BASMAT>

RATIONAL TIME RESPONSE - <RIRESP>

STATE VARIABLE FEEDBACK - <5TVAR>

CONTROLLABILITY AND OBSERVABILITY - <OBSCON>

LUENBERGER OBSERVER - <LUEN>

OPTIMAL CONTROL/KALMAN FILTER - <KALMAN>

OPTIMAL CONTROL - <OPICON>

PARTIAL FRACTION EXPANSION - <PRFEXP>

ROOTS OF A POLYNOMIAL - <ROOTS>

TO USE ONE OF THE SUBPROGRAMS ENTER THE NAME BETWEEN THE SYMBOLS <>. rccts

ROOTS IS USED TO PIND THE ROOTS OF A POLYNOMIAL OF DEGREE LESS THAN OR EQUAL TO EIGHT.

FIRST ENTER THE PROBLEM IDENTIFICATION (NOT TO EXCEED 20 CHARACTERS). thesis example

ENTER THE ORDER OF THE SYSTEM (UP TO 8).

ROOTS OF A POLYNOMIAL COEFFICIENTS IN ASCENDING ORDER OF S.

\*\*WARNING--THE HIGHEST ORDER COEFFICIENT MUST BE UNITY.\*
COEFF(1)=

COEFF(2)=

5.1

 $_{3}$  COEFF(3)=

COEFF(4)=

POLYNOMIAL COEFFICIENTS - IN ASCENDING POWERS OF S 8.00E-01 5.10E+00 3.33E+00 1.03E+00

THE ROOTS ARE REAL PART IMAG. PART -1.41E+00 1.62E+00 -1.41E+00 -1.62E+00 -1.74E-01 0.0

THIS CONCLUDES THE POLYNOMIAL ROOTS PROGRAM.

DO YOU WANT TO RUN THE PROGRAM AGAIN?

ANALYSIS IS COMPLETE. DO YOU WANT TO RUN LINCON AGAIN?  $\boldsymbol{n}$ 

LINCON IS NOW TERMINATED.

The results shown in Appendix C are easily

interpreted.

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### E. RATIONAL TIME RESPONSE PROGRAM (RTRESP)

The time response in closed form of a linear control system described by Eqs. (3.A-1), (3.A-2) and (3.A-3) is calculated by this program. The user must define the initial conditions  $\mathbf{x}(3)$  and the rational Laplace transform of the scalar forzing function  $\mathbf{r}(t)$ . The theoretical concepts involved in the development of the computer codes are described by Melsa [1].

1. Input Restrictions And Limitations

Enter the following in F format (floating point):

- 1) the elements of the plant matrix A
- 2) the control vector b
- 3) the output vector g
- 4) the feed backback coefficient vector k
- 5) the controller gain K

and

6) the initial conditions vector  $\mathbf{x}(0)$ .

The rational Laplace transform of the input function must be in the form

$$Z [r(t)] = R(s) = G \frac{N(s)}{D(s)}$$
where  $G = \text{the input function gain,}$ 

$$N(s) = n_1 + n_2 s + n_3 s^2 + \dots + s^3$$
and  $D(s) = d_1 + d_2 s + d_3 s^2 + \dots + s^5$ 

The numerator and denominator coefficients must be arranged so that the coefficients of  $s^3$  and  $s^4$  are each unity with  $p>q\geq 0$ . Due to programming limitations it is necessary that the order of the system plus the dimension of D(s) be less than or equal to ten.

Upon entering the input function gain, the user next has the option to enter the numerator and denominator in either polynomial coefficient form or factored root form. With the factored root form enter the real part of the root as negative if it lies in the left half plane and just the magnitude of the imaginary part.

### 2. Terminal Sassion Exampla

This section contains a terminal session for a specific example. Commands entered by the user are in lower case. The RTRESP program will determine the time response form of the closed-loop system

$$\underline{\mathbf{x}}(t) = \begin{bmatrix} 1.0 & 0.0 & 0.0 \\ 0.0 & -2.3 & 1.0 \\ 0.0 & -0.5 & 1.0 \end{bmatrix} \underline{\mathbf{x}}(t) + \begin{bmatrix} 0.0 \\ 0.0 \\ 1.0 \end{bmatrix} \mathbf{u}(t)$$

$$u(t) = 3.2\{r(t) - [1.0 1.0 0.0] \underline{x}(t)\}$$

$$\mathbf{y}(t) = [1.0 \ 0.4 \ 1.0] \ \mathbf{x}(t)$$

if the Laplace transform of the input function is

$$R(s) = 0.5 \quad \frac{s + 1.0}{s^2 - 2.0}$$

and the initial conditions are zero.

Note here that the open-loop rational time response may be calculated by setting the feedback coefficient vector Also, if only an initial k equal to a zero vector. condition response is desired, the input function gain G is set to zero.

### lincon

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EXECUTION BEGINS...

LINCON CONSISTS OF THE FOLLOWING SUBPROGRAMS:

BASIC MATRIX MANIPULATION - <BASMAT>

RATIONAL TIME RESPONSE - <RIRESP>

STATE VARIABLE FEEDBACK - <STVAR>

CONTROL LABILITY AND OBSERVABILITY - <OBSCON>

LUENBERGER OBSERVER - <LUEN>

OPTIMAL CONTROL/KALMAN FILTERS - <RICATI>

DISCRETE TIME KALMAN FILTER - <KALMAN>

OPTIMAL CONTROL - <CPTCON>

PARTIAL FRACTION EXPANSION - <PRPEXP>

ROOTS OF A POLYNOMIAL - <ROOTS>

TO USE ONE OP THE SJB PROGRAMS ENTER THE NAME BETWEEN THE SYMBOLS < >.

ETLESP rtresp

RTRESP DETERMINES THE TIME RESPONSE OF A LINEAR FEEDBACK CONTROL SYSTEM. THIS PROGRAM WILL GIVE A CLOSED-FORM EXPRESSION FOR THE TIME RESPONSE.

FIRST ENTER THE PROBLEM IDENTIFICATION (NOT TO EXCEED 20 CHARACTERS). thesis example

ENTER THE ORDER OF THE SYSTEM (UP TO 8).

ENTER THE ELEMENIS OF THE PLANT MATRIX -- A.

1.04(1,1)=

0.0 A (1, 2) =

0.0<sup>A</sup>(1,3) =

0.0 A (2, 1) =

 $-2.0^{A(2,2)}$  =

1.0 4 (2, 3) =

0.0<sup>A</sup>(3,1)=

 $-0.5^{(3,2)}$  =

1.0<sup>A</sup>(3,3) =

THE A MAIRIX (PLANT MATRIX)

1.00E+0G 0.0 -2.00E+00 1.30E+00 0.0 -5.30E-01 1.30E+00

DO YOU WANT TO CHANGE ANY ELEMENT (S) OF THE MATRIX?

ENTER THE ELEMENTS OF THE CONTROL VECTOR--B.

0.08(1)=

0.03(2)=

1.0<sup>B(3)=</sup>

THE B MATRIX (CONTROL VECTOR)

0.0 0.0 1.COE+00

DO YOU WANT TO CHANGE ANY ELEMENT(S) OF THE MATRIX?

```
ENTER THE ELEMENIS OF THE DUIPUT VECTOR -- C.
1.0 (1) =
0.4<sup>C</sup>(2)=
1.0<sup>C(3)=</sup>
     THE C MATRIX (OUTPUT VECTOR)
   DO YOU WANT TO CHANGE ANY ELEMENT (S) OF THE MATRIX?
   ENTER THE ELEMENIS OF THE FEEDBACK COEFFICIENT VECTOR -- FDBG.
1.0 FDBG(1) =
1.0 FDBG (2) =
0.0 FDBG (3) =
     THE FEEDBACK COEFFICIENT VECTOR
         1.00E+00
1.00E+00
0.0
    DO YOU WANT TO CHANGE ANY ELEMENT(S) OF THE VECTOR?
ENTER THE CONTROLLER GAIN--K.
     THE CONTROLLER GAIN
         3.20E+00
   DO YOU WANT TO CHANGE THE VALUE OF THE GAIN?
   ENTER THE ELEMENTS OF THE INITIAL CONDITIONS VECTOR -- \mathbf{X}\left(0\right).
0.0 %0 (1) =
J. 0 XO (2) =
0.0 XO (3) =
     INITIAL CONDITIONS VECTOR-X (3)
```

DO YOU WANT TO CHANGE ANY ELEMENT (S) OF THE VECTOR?

ENTER THE GAIN FOR THE RATIONAL LAPLACE TRANSFORM OF THE INPUT FUNCTION.

THE INPUT FUNCTION GAIN = 5.00E-01

DO YOU WANT TO CHANGE THE VALUE OF THE GAIN?

ENTER THE NUMERALOR BY POLYNOMIAL COEFFICIENT OR FACTORED ROOF FORM. FIRST ENTER EITHER THE LETTER P FOR POLYNOMIAL COEFFICIENT FORM OR THE LETTER P POR FACTORED ROOT FORM.

ENTER THE NUMERATOR POLYNOMIAL ORDER.

ENTER THE REAL PART OF THE ROOT.

ENTER THE MAGNITIDE OF THE IMAGINARY ROOT.

NUMERATOR POLYNOMIAL OF R(S) -ASCENDING POWERS OF S 1.00E+00 1.00E+00

NUMERATOR ROOIS ARE REAL PART IMAG. PART -1.00E+00 0.3

ENTER THE DENOMINATOR BY POLYNOMIAL COEFFICIENT OR FACTORED ROOF FORM. FIRST ENTER EITHER THE LETTER P FOR POLYNOMIAL COEFFICIENT FORM OR THE LETTER P FOR FACTORED ROOT FORM.

ENTER THE DENOMINATOR POLYNOMIAL ORDER.

THE POLYNOMIAL COEFFICIENTS MUST BE ENTERED IN ASCENDING ORDER OF S.

WARNING -- THE HIGHEST ORDER COEFFICIENT MUST BE UNITY. DO YOU NEED TO CHANGE THE INPUT FUNCTION GAIN TO SATISFY THIS REQUIREMENT?

2NTER THE POLYNOMIAL COEFFICIENTS IN
ASCENDING ORDER OF S.
CO (1) =
-2.3

0.0 (2) =

P

1.0 (3) =

DENOMINATOR POLYNOMIAL OF R(S) - ASCENDING POWERS OF S
-2.00E+00 0.0 1.00E+00

DENOMINATOR ROOTS ARE REAL PART IN AG. PART 1.41E+00 0.3 -1.41E+00 0.3

THE TIME RESPONSE OF THE STATE X (T)

THE VECTOR COEFFICIENT OF EXP(-5.0E-01) T\*COS(1.2E+00) T

0.0 -3.70E-01 -9.72E-01

THE VECTOR COEFFICIENT OF EXP (-5.0E-01) T\*SIN (1.2E+00) T

0.0 -3.47E-01 -7.50E-02

THE VECTOR COEFFICIENT OF EXP( 1.00E+00) T

0.0 0.3 2.58E-06

THE VECTOR COEFFICIENT OF EXP ( 1.41E+00) T

0.0 2.67E-01 9.12E-01

THE VECTOR COEFFICIENT OF EXP( -1.41E+00) T

0.0 1.03E-01 5.00E-02

\*\*\*\*\*\*\*\*\*\*\*\*\*

THE TIME RESPONSE OF THE OUTPUT Y (T)

THE COEFFICIENT OF EXP( -5.00E-01) T\*COS( 1.20E+00) T

-1. 12E+00

THE COEFFICIENT OF EXP( -5.00E-01) T\*SIN( 1.20E+00) T

-2. 14E-01

THE COEFFICIENT OF EXP ( 1.00E+00) T

2.68E-06

THE COEFFICIENT OF EXP( 1.41E+00)T

1.02E+00

THE COEFFICIENT OF EXP( -1.41E+00) T

1.01E-01

ANALYSIS IS COMPLETE. DO YOU WANT TO RUN LINCON AGAIN?

n

LINCON IS NOW TERMINATED.

The computer results shown in Appendix D are interpreted as:

$$x_1(t) = 0.0$$

$$x_1(t) = -0.37 \exp(-0.5t) \cos(1.2t) - 0.35 \exp(-0.5t) \sin(1.2t) + 0.67 \exp(1.41t) + 0.13 \exp(-1.41t)$$

$$x_3(t) = -0.97 \exp(-0.5t) \cos(1.2t) - 0.08 \exp(-0.5t) \sin(1.2t) + 2.7x10^{-6} \exp(t) + 0.91 \exp(1.41t) + 5.06 \exp(-1.41t)$$

$$y(t) = -1.12 \exp(-0.5t) \cos(1.2t) -0.21 \exp(-0.5t) \sin(1.2t) +2.7x10^{-6} \exp(t) +1.02 \exp(1.41t) +0.1 \exp(-1.41t)$$

## IV. MODERN CONTROL PROGRAMS

#### A. INTRODUCTION

In this chapter six programs are presented and discussed which may be used for the analysis and design of control systems.

The first, OBSCON, is used to find the observability index and controllability of a system. The next two programs are used to design Kalman filters. RICATI and KALMAN can find the feedback and control gains necessary to optimize a function for either continuous or discrete systems, respectively. The last three programs may be used to design optimal linear control systems. STVAR is particularly useful in the design of linear state variable feedback control systems. It may be used to calculate both open- and closed-loop transfer functions and also has the ability to design a closed-loop system from desired transfer function specifications. LUEN is used to design a combined, reduced-order, observer-controller to achieve a desired closed-loop transfer function from a system where some of the states are inaccesible. OPTCON will minimize a given cost function producing a scalar control.

B. OBSERVABILITY INDEX AND CONTROLLABILITY PROGRAM (OBSCON)

This program is used to determine the observability index and controllability of the linear system

$$\underline{\dot{x}}(t) = \underline{\lambda}\underline{x}(t) + \underline{B}\underline{u}(t) \tag{4.8-1}$$

$$\underline{\mathbf{Y}}(t) = \underline{\mathbf{C}}\underline{\mathbf{X}}(t) \tag{4.8-2}$$

where

x =state vector (n-dimensional vector)

u = control vector (1-dimensional vector)

y = output vector (m-dimensional vector)

 $\underline{\mathbf{A}} = \mathbf{n} \times \mathbf{n}$  matrix

 $B = n \times 1$  matrix

 $C = m \times n \text{ matrix}$ 

The observability index is defined as the minimum integer such that the matrix

$$[\underline{C},\underline{A}^{\mathsf{T}}\underline{C},\ldots,(\underline{A}^{\mathsf{T}})^{r-1}\underline{C}]$$

has rank n. The above system is said to be controllable at a given initial time if it is possible, by using an

unconstrained control vector, to force the system from an initial state of  $\mathbf{x}(0)$  to some other state in a finite time interval [4].

If the user desires just the observability index to be calculated, enter the B matrix as a zero matrix. Likewise, if just the controllability of the system is desired, enter the C matrix as a zero matrix. The unobservable or uncontrollable system response is then of course ignored.

## 1. Input Restrictions

Other than the limitations to the problem identification, (20 characters) and system size (8x8) there are no restrictions to OBSCON. A reminder here may be helpful, however. Remember to enter your elements of the matrices in F format.

## 2. Terminal Session Example

This section contains a session for a specific problem. Commands entered by the user are in lower case. The following system is to be tested:

$$\underline{\mathbf{x}}(t) = \begin{bmatrix} 0.0 & -1.0 & 0.0 \\ -1.0 & -0.5 & 1.0 \\ 0.0 & 0.0 \end{bmatrix} \underline{\mathbf{x}}(t) + \begin{bmatrix} 2.0 & 1.0 \\ 0.0 & 1.0 \\ 0.0 & 0.0 \end{bmatrix} \underline{\mathbf{u}}(t)$$

```
\mathbf{Y}(t) = \begin{bmatrix} 0.0 & 1.0 & 0.0 \\ 1.0 & 1.0 & 0.0 \\ -2.0 & 1.0 & 0.0 \end{bmatrix} \mathbf{X}(t)
```

lincon

EXECUTION BEGINS... THE FOLLOWING SUBPROGRAMS:

BASIC MATRIX MANIPULATION - <BASMAT>
RATIONAL TIME RESPONSE - <RIRESP>
STATE VARIABLE FEEDBACK - <SIVAR>
CONTROL LABILITY AND OBSERVABILITY - <OBSCON>
LUENBERGER OBSERVER - <LUEN>
OPTIMAL CONTROL/KALMAN FILTERS - <RICATI>
DISCRETE TIME KALMAN FILTER - <KALMAN>
OPTIMAL CONTROL - <OPICON>
PARTIAL FRACTION EXPANSION - <PREXP>
ROOTS OF A POLYNOMIAL - <ROOTS>
TO USE ONE OF THE SUBPROGRAMS ENTER THE NAME BETWEEN THE SYMBOLS <>.
Obscon

OBSCON DETERMINES THE OBSERVABILITY INDEX AND CONTROLIABILITY OF A SYSTEM.

FIRST ENTER THE PROBLEM IDENTIFICATION (\*NOT TO EXCEED 2) CHARACTERS\*) thesis example

NOW, ENTER THE ORDER OF THE SYSTEM (UP TO 8).

ENTER THE ELEMENIS OF THE PLANT MATRIX -- A.

$$-1.0^{A(2,1)}$$

$$-0.5^{A(2,2)} =$$

```
THE A MATRIX (PLANT MATRIX)
0.0 1.00E+00 0.0
-1.00E+00 -5.00E-01 1.00E+00
0.0 1.00E+00
```

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

ENTER THE NUMBER OF COLUMNS OF THE B MATRIX.

ENTER THE ELEMENTS OF THE B MATRIX.

B (1, 1) =

THE B MATRIX

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

ENTER THE NUMBER OF OUTPUTS.

ENTER THE ELEMENTS OF THE C MATRIX. G.O

$$-2.0^{\circ}(3,1) =$$

THE C MATRIX

0.0 1.00E+00 -2.00E+00 1.00E+00 1.00E+00 0.0

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

OBSERVABILITY INDEX = 2

THE SYSTEM (A, B) IS UNCONTROLLABLE

n

THIS CONCLUDES THE OBSERVABILITY INDEX/CONTROLLABILITY PROGRAM (OBSCON)

ANALYSIS IS COMPLETE. DO YOJ WANT TO RUN LINCON AGAIN?

LINCON IS NOW TERMINATED.

The complete results presented in Appendix E should be self-explanatory to the user.

## C. OPTIMAL CONTROL/KALMAN FILTER PROGRAM (RICATI)

The transient solution, matrix gains, to the Riccati differential equations for the state-regulator controller and the continuous Kalman filter are calculated by the RICATI program.

Given a state-regulator problem with the linear, timeinvariant system

$$\frac{\dot{\mathbf{x}}}{\dot{\mathbf{x}}}(t) = \underline{\mathbf{A}}\underline{\mathbf{x}}(t) + \underline{\mathbf{B}}\underline{\mathbf{u}}(t) \tag{4.C-1}$$

$$\mathbf{Y}(t) = \mathbf{C}\mathbf{X}(t) \tag{4.C-2}$$

the Riccati equation is defined as

$$\underline{\hat{P}}(t) = -\underline{P}(t)\underline{A} - \underline{A}^{\mathsf{T}}\underline{P}(t) + \underline{P}(t)\underline{B}\underline{R}^{-1}\underline{B}^{\mathsf{T}}\underline{P}(t) - \underline{Q}$$
 (4. C-3)

with  $\underline{P}(t_0)$  as the boundary condition. If  $\underline{u}(t)$  is not constrained, a gain matrix can be found such that the cost function

$$J=1/2\left(\underline{x}^{\mathsf{T}}(t_{\mathsf{f}})\underline{p}(t_{\mathsf{f}})\underline{x}(t_{\mathsf{f}})\right)+1/2\int_{t_{\mathsf{g}}}^{t_{\mathsf{f}}}\left(\underline{x}^{\mathsf{T}}(t)\underline{Q}\underline{x}(t)+\underline{u}^{\mathsf{T}}(t)\underline{R}\underline{u}(t)\right)dt(4.6-4)$$

is minimized [ 5]. Such a gain matrix is defined as

$$\underline{G}_{c}(t) = \underline{R}^{-1}\underline{B}^{\mathsf{T}}\underline{P}(t) \tag{4.6-5}$$

and

$$\underline{\mathbf{u}}(t) = -\underline{\mathbf{G}}_{c}(t)\underline{\mathbf{x}}(t) \tag{4.C-6}$$

The transient solution is solved by the computer. Note that the output from the computer for the gain matrix does not include the negative sign of the feedback loop.

Given a continous Kalman filter problem with the linear, time-invariant system

$$\underline{x}(t) = \underline{\lambda}\underline{x}(t) + \underline{\beta}\underline{y}(t) \qquad (4.C-7)$$

$$\underline{\mathbf{Y}}(t) = \underline{\mathbf{C}}\underline{\mathbf{X}}(t) + \underline{\mathbf{Y}}(t) \tag{4. C-8}$$

where  $\underline{w}(t)$ , the random process forcing input and  $\underline{v}(t)$ , the measurement noise, have covariance matrices of  $\underline{R}$ , the random input covariance matrix, and  $\underline{Q}$ , the measurement noise covariance matrix, respectfully, the Riccati equation is defined as

$$\dot{\underline{P}}(t) = \underline{A}\underline{P}(t) + \underline{P}(t)\underline{A} + \underline{B}\underline{Z}\underline{B}^{T} - \underline{P}(t)\underline{C}^{T}\underline{R}^{-1}\underline{C}\underline{P}(t)$$
 (4. C-9)

with

$$\underline{\underline{p}}(t_0) = E[[\underline{\hat{x}}(t_0) - \underline{x}]] \underline{\hat{x}}(t_0) - \underline{x}]^T$$
 (4.2-10)

as the initial condition boundary. The gain matrix found is defined as

$$\underline{G}_{L}(\tau) = \underline{R}^{-1}\underline{CP}(\tau) \tag{4.2-11}$$

## 1. Terminal Session Example

This section contains a session for the specific example presented by Melsa [ 1]. Commands entered by the user are in lower case. Given the second order linear system

$$\underline{\hat{\mathbf{x}}}(t) = \begin{bmatrix} -1.0 & 0.0 \\ 0.5 & 0.0 \end{bmatrix} \underline{\mathbf{x}}(t) + \begin{bmatrix} 1.0 & 0.0 \\ 0.0 & 1.0 \end{bmatrix} \underline{\mathbf{u}}(t)$$

$$\underline{\mathbf{x}}(t) = \begin{bmatrix} 1.0 & 0.0 \\ 0.0 & 2.0 \end{bmatrix} \underline{\mathbf{x}}(t)$$

determine the optimal transient response control and filter gains for

$$\underline{Q} = \overline{\underline{Q}} = \begin{bmatrix} 1.0 & 1.0 \\ 1.0 & 1.0 \end{bmatrix}$$

$$\underline{R} = \overline{\underline{R}} = \begin{bmatrix} 1.0 & 0.0 \\ 0.0 & 2.0 \end{bmatrix}$$

For the control option an initial time of 0.0 and a final time of 10.0 is used. For the filter option an initial time of 0.0 and a final time of 5.0 is used. Also, for the filter option, the initial condition matrix is chosen to be

$$P(+_{o}) = \begin{bmatrix} 0.0 & 0.0 \\ 0.0 & 0.0 \end{bmatrix}$$

For both options ten equally spaced values of G and G on the time interval  $t \le t \le t$  are used, i.e., NPOINT is set equal to 10.

lincon

EXECUTION BEGINS...

LINCON CONSISTS OF THE POLLOWING SUBPROGRAMS:

BASIC MATRIX MANIPULATION - <BASMAT>
RATIONAL TIME RESPONSE - <RIRESP>
STATE VARIABLE FEEDBACK - <SIVAR>
CONTROL LABILITY AND OBSERVABILITY - <OBSCON>
LUENBERGER OBSERVER - <LUEN>
OPTIMAL CONTROL/KALMAN FILTERS - <RICATI>
DISCRETE TIME KALMAN FILTER - <KALMAN>
OPTIMAL CONTROL - <OPTION>
PARTIAL FRACTION EXPANSION - <PRFEXP>
ROOTS OF A POLYNOMIAL - <ROOTS>
TO USE ONE OF THE SUBPROGRAMS ENTER THE NAME BETWEEN THE SYMBOLS < >.
LICATI

RICATI DETERMINES THE TRANSIENT SOLUTION FOR THE RICCATI EQUATION. FIRST ENTER THE PROBLEM IDENTIFICATION (\*NOT TO EXCEED 2) CHARACTERS\*) thesis example

NOW, ENTER THE ORDER OF THE SYSTEM (UP TO 8).

ENTER THE ELEMENTS OF THE PLANT MATRIX -- A.

$$-1.0^{\frac{1}{0}(1,1)}$$
=

0.0<sup>A</sup>(1,2) =

0.0<sup>A</sup>(2,1)=

 $-2.0^{A(2,2)}$ =

-1.00E+00 J.0 -2.00E+00

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

ENTER THE NUMBER OF CONTROL INPUTS.  $\mathbf{2}$ 

```
ENTER THE ELEMENTS OF THE DISTRIBUTION MATRIX -- B.
```

1.08(1,1)=

0.0<sup>B(1,2)=</sup>

0.0<sup>B</sup>(2.1) =

1.08(2,2)=

THE B MATRIX (DISTRIBUTION MATRIX)

1.00E+00 0.0 1.00E+00

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

ENTER THE NUMBER OF OBSERVABLE OUTPUTS.

ENTER THE ELEMENIS OF THE MEASUREMENT MATRIX -- C.

1.0<sup>C</sup>(1,1)=

0.0<sup>C</sup>(1,2)=

0.0<sup>C</sup>(2,1) =

2.0<sup>C</sup>(2,2) =

THE C MAIRIK (MEASUREMENT MAIRIX)

1.00E+00 0.0 2.00E+00

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

YOU HAVE IWO OPTIONS AVAILABLE:

(1) THE CONIROL OPTION FOR SOLVING STATE-REGULATOR

PROBLEMS, OR

(2) THE FILTER OPTION FOR SOLVING A CONTINUOUS KALMAN

PILTER PROBLEM.

FOR THE CONTROL OPTION, ENTER THE LETTER C.

FOR THE FILTER OPTION, ENTER THE LETTER F.

ENTER THE ELEMENTS OF THE CONTROL WEIGHTING MATRIX -- R.

1.0 R (1, 1) =

0.0<sup>R</sup>(1,2)=

0.03(2,1)=

2.0<sup>R</sup>(2,2)=

THE R MATRIX (CONTROL WEIGHTING MATRIX)

1.00E+00 0.0 0.0 2.00E+00

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

ENTER THE ELEMENTS OF THE STATE WEIGHTING MATRIX--Q.

1.0 0 (1, 1) =

1.0 (1,2) =

1.0 (2,1) =

1.0 (2,2) =

THE Q MATRIX (STATE WEIGHTING MATRIX)

1.00E+00 1.00E+00 1.00E+00 1.00E+00

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

ENTER THE INITIAL TIME FOR THE TRANSIENT RESPONSE.

ENTER THE FINAL TIME FOR THE TRANSIENT RESPONSE.

ENTER THE NUMBER OF POINTS OF THE TRANSIENT RESPONSE TO BE PRINTED. (<100)

ENTER THE ELEMENTS OF THE PERMINAL BOUNDARY VALUE MATRIX -- P.

0.0<sup>P(1,1)=</sup>

0.0 (1,2) =

0.0 2 (2, 1) =

0.0<sup>P(2,2)=</sup>

THE P MATRIX (TERMINAL BOUNDARY VALUE MATRIX)

0.3

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

#### TRANSIENT SOLUTION

T T T II T	INTRALIBUT GODGILG	
TIME = GAINS 0.0 0.0	1.000E+01	
TIME = GAINS	9.000E+00	
3.75E-01	2.81E-01	
1.40E-01	1.11E-01	
TIME = GAINS	8.000E+00	
3.98E-01	2.85E-01	
1.42E-01	1.12E-01	
TIME = GAINS	7.000E+03	
4.00E-01	2.85E-01	
1.42E-01	1.12E-01	
TIME = GAINS	6.000E+03	
4.00E-01	2.85E-01	
1.42E-01	1.12E-01	
TIME = GAINS	5.000E+03	
4.00E-01	2.85E-01	
1.42E-01	1.12E-01	
TIME = GAINS	4.000E+03	
4.002-01	2.85E-01	
1.42E-01	1.12E-31	
TIME = GAINS 4.00E-01 1.42E-01	3.000E+03 2.85E-01 1.12E-01	
TIME = GAINS	2.000E+03	
4.00E-01	2.85E-01	
1.42E-01	1.12E-31	
TIME = GAINS	1.000E+03	
4.00E-01	2.85E-01	
1.42E-01	1.12E-01	
TIME = GAINS 4.00E-01	2.861E-06 2.85E-01 1.12E-01	

```
DO YOU WANT THE FILTER OPTION?
```

WILL THERE BE CHANGES TO THE A, B, OR C MATRICES?

WILL THERE BE CHANGES TO THE R OR Q MATRICES?

ENTER THE INITIAL TIME FOR THE TRANSIENT RESPONSE.

ENTER THE FINAL FIME FOR THE TRANSIENT RESPONSE.

ENTER THE NUMBER OF POINTS OF THE TRANSIENT RESPONSE TO BE PRINTED. (<100)

\*\*\*\*\*\*\*\*\*\*\*\*\*\*\*\*\*\*\* \*\*\* PILTER OPTION \*\*\* \*\*\*\*\*\*\*\*\*\*\*\*\*\*

ENTER THE ELEMENTS OF THE INITIAL BOUNDARY VALUE MATRIX -- P.

0.0<sup>P(1,1)=</sup>

0.0<sup>P(1,2)=</sup>

0.0<sup>P</sup>(2,1) =

0.0<sup>P</sup>(2,2)=

THE P MATRIX (INITIAL BOUNDARY VALUE MATRIX)

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

TRANSIENT SOLUTION

TIME = 0.0 GAINS

0.0

TIME = 5.000E-01

GAINS 2.39E-01 2.00E-01 6.02E-01 6.85E-01

TIME = 1.000E+03

GAINS 2.71E-01 2.13E-01 8.51E-01 7.05E-01

TIME = 1.500E+00

GAINS 2.77E-01 2.13E-01 8.53E-01 7.36E-01

```
TIME =
                2.000E+00
GAINS
2.78E-01
8.53E-01
   TIME =
                2.500E+00
GAINS
2.79E-01
8.53E-01
TIME = GAINS
2.79E-01
8.53E-01
                3.000E+00
   TIME =
                3.500E+00
   TIME =
                4.000E+03
TIME = GAINS
2.79E-01
8.53E-01
                4.500E+00
               2.13E-01
7.06E-01
   TIME =
                5.000E+00
```

\*\*\*\*\*\*\*\*\*\*\*\*\*\*\*\*\*\*

SUCITED TORINCE BHL THEM OF OF

THIS CONCLUDES THE RICCATI EQUATION PROGRAM (RICATI).

ANALYSIS IS COMPLETE. DO YOU WANT TO RUN LINCON AGAIN?

LINCON IS NOW TERMINATED.

The computer results for both options presented in Appendix F indicates that a steady state gain matrix has been achieved.

## D. DISCRETE TIME KALMAN FILTER (KALMAN)

The discrete Kalman filter gain matrix, G(k), is calculated by this program. The theoretical concepts involved in the development of the computer codes are described by Sage [5]. A brief development of the discrete Kalman filter is included here as an aid in the use of KALMAN. Additional reference materials can be found in the bibliography.

Kalman filtering is a method of obtaining minimum-variance estimates of signals from noisy measurements. The discrete Kalman filter provides state estimates for the following system

$$\underline{\mathbf{x}}(\mathbf{k}) = \underline{\underline{\Phi}}\underline{\mathbf{x}}(\mathbf{k}-1) + \underline{\underline{\Delta}}\underline{\mathbf{u}}(\mathbf{k}-1) + \underline{\underline{\Gamma}}\underline{\mathbf{v}}(\mathbf{k}-1)$$
 (4. D-1)

with the discrete linear observations

$$\underline{\mathbf{z}}(\mathbf{k}) = \underline{\mathbf{H}}\underline{\mathbf{x}}(\mathbf{k}) + \underline{\mathbf{v}}(\mathbf{k}) \tag{4. D-2}$$

where

x =the nx1 state vector at the time t(k)

 $\Phi$  = the nxn nonsingular state transition matrix

 $\Gamma$  = the nxr disturbance transition or distribution matrix

 $\Delta$  = the nxp control distribution matrix

w = the rx1 disturbance of system random input vector

z = the mx1 measurement vector

 $\underline{\mathbf{H}}$  = the mx1 measurement or observation vector

▼ = the mx1 measurement noise vector

g = the px1 control or test signal vector

k =the discrete-time index (k = 0, 1, ...)

The optimal filtered estimate of  $\underline{x}(k)$  , denoted  $\hat{\underline{x}}(k|k)$  , is given by the recursive relations

$$\hat{\mathbf{x}}(k|k-1) = \underline{\Phi}\mathbf{x}(k-1|k-1) + \underline{\Psi}\mathbf{u}(k-1)$$
 (4.D-3)

an d

for  $k = 0, 1, \dots$ , where  $\underline{x}(0, 0) = \underline{0}$ . The Kalman gain matrix,  $\underline{G}(k)$ , is an nxm matrix which is specified as

$$\underline{G}(k) = \underline{P}(k \mid k-1) \underline{H}^{T} [\underline{H}\underline{P}(k \mid k-1) \underline{H}^{T} + \underline{R}(k)] - 1 \qquad (4. D-5)$$

$$\underline{P}(k|k) = [\underline{I} - \underline{G}(k) \underline{H}] \underline{P}(k|k-1)$$
 (4.0-6)

$$P(k|k-1) = \Phi P(k-1|k-1) \Phi^{T} + 2(k,k-1)$$
 (4. D-7)

where

I = the nxn identity matrix

P(0|0) = P(0), the initial condition matrix

 $\hat{\mathbf{x}}(k|k-1)$  = the single-stage prediction of  $\mathbf{x}(k)$ 

 $\hat{\mathbf{x}}(\mathbf{k}|\mathbf{k})$  = the filtered estimate of  $\mathbf{x}(\mathbf{k})$ 

 $Q = \left[\mathbb{E}\left\{\underline{w}\left(k\right)\underline{w}\left(k\right)^{\mathsf{T}}\right\}\right]^{\mathsf{T}}$ , the nxn covariance matrix of the random input

R = the mxn covariance matrix of the measurement noise

 $E\{\underline{w}(k)\underline{w}(k)^{\top}\}$ , the mean-square magnitude of the perturbation acceleration matrix

## 1. Input Requirements

This program computes the recurrence Eqs. (4.D-5), (4.D-6) and (4.D-7) for a specified number of iterations NP and prints 3(k) as a function of k.

The required inputs are

- 1) the transition matrix  $(\frac{\pi}{2})$
- 2) the distribution matrix  $(\Gamma)$
- 3)  $\mathbb{E}\{\underline{\mathbf{w}}(\mathbf{k})\underline{\mathbf{w}}(\mathbf{k})^{\top}\}$
- 4) the observation matrix  $(\underline{H})$
- 5) the number of points to be performed (NP)

and

6) the initial condition matrix [P(0|0)].

#### 2. Terminal Session Example

This section contains a session for a specific problem. Commands entered by the user are in lower case. The following system is to be tested:

$$\underline{\mathbf{x}}(k) = \begin{bmatrix} 1.0 & 0.5 \\ 0.0 & 1.0 \end{bmatrix} \underline{\mathbf{x}}(k-1) + \begin{bmatrix} 0.125 \\ 0.5 \end{bmatrix} \underline{\mathbf{w}}(k-1)$$

$$\underline{z}(k) = [1.0 \ 0.0] \underline{x}(k) + \underline{y}(k)$$

 $\underline{\mathbf{R}} = [5.0]$ 

 $E\{\underline{w}(k)\underline{w}(k)^{\mathsf{T}}\} = 4.0$ 

$$\underline{P}(0) = \begin{bmatrix} 1000.0 & 0.0 \\ 0.0 & 1000.0 \end{bmatrix}$$

The number of time points to be computed is chosen to be 20.

lincon

EXECUTION BEGINS. FIHE FOLLOWING SUBPROGRAMS:

LINCON CONSISTS OF THE FOLLOWING SUBPROGRAMS:

BASIC MATRIX MANIPULATION - <BASMAT>
RATIONAL TIME RESPONSE - <RIRESP>

STATE VARIABLE FEEDBACK - <STVAR>
CONTROLIABILITY AND OBSERVABILITY - <OBSCON>
LUENBERGER OBSERVER - <LUEN>
OPTIMAL CONTROL/KALMAN FILTERS - <RICATI>
DISCRETE TIME KALMAN FILTER - <KALMAN>
OPTIMAL CONTROL - <OPICON>
PARTIAL FRACTION EXPANSION - <PREXP>
ROOTS OF A POLYNOMIAL - <ROOTS>
TO USE ONE OF THE SUBPROGRAMS ENTER THE NAME BETWEEN THE SYMBOLS <>.

Kalman

KALMAN DETERMINES THE DISCRETE KALMAN FILTER GAIN MATRIX --G(K). FIRST ENTER THE PROBLEM IDENTIFICATION (\*NOT TO EXCEED 2) CHARACTERS\*). thesis example

NOW, ENTER THE ORDER OF THE SYSTEM (UP TO 8).

```
ENTER THE ELEMENIS OF THE TRANSITON MATRIX--PHI PHI(1,1)=
```

PHI(1,2)=

0.0 PHI (2, 1)=

PHI(2,2)=

THE PHI MATRIX (PRANSITION MATRIX)

1.00E+00 0.0 5.03E-01 1.00E+00

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

ENTER THE DIMENSION OF THE RANDOM INPUT VECTOR.

ENTER THE ELEMENTS OF THE DISTRIBUTION MATRIX--GAMMA. GAMMA(1,1) = .125

.5 GAMMA (2,1) =

THE GAMMA MATRIX (DISTRIBUTION MATRIX)

1. 25 E-01

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

ENTER THE ELEMENTS OF THE MEAN-SQUARE MAGNITUDE OF THE PERTURBATION ACCELERATION MATRIX--W.

W (1, 1) =

THE W MATRIX (MEAN-SQUARE MAGNITUDE OF THE PERTUR BATION ACCELERATION MAIRIX)

4.00E+00

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

ENTER THE NUMBER OF OUTPUTS.

ENTER THE ELEMENTS OF THE OBSERVATION MATRIX -- H.

H(1,1) =

H(1,2) = 0.

THE H MATRIX (OBSERVATION MATRIX)

1.00E+00 0.0

```
DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
```

ENTER THE ELEMENTS OF THE MEASUREMENT NOISE COVARIANCE MATRIX -- R.

R(1,1) =

THE R MATRIX (MEASUREMENT NOISE COVARIANCE MATRIX) 5.00 E+00

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

ENTER THE NUMBER OF THE POINTS TO BE PERFORMED. (<100)

---

ENTER THE ELEMENIS OF THE INITIAL CONDITION MATRIX -- P.

P(1,1) =

P(1,2) =

P(2,1) =

p(2,2) = 1000.

THE P MATRIX (INITIAL CONDITION MATRIX)

K = 0 GAINS 9.95E-01 0.

9.95E-01 0.0

K = 1
GAINS
9.81E-01 1.92E+00

K = 2 GAINS 8.29E-01 9.93E-01

K = 3 GAINS 7.03E-01 6.24E-01

K = 4 GAINS 6.13E-01 4.52E-01

K = 5 GAINS 5.54E-01	3.72E-01
K = 6 GAINS 5.18E-01	3.35E-01
K = 7 GAINS 4.99E-01	3.23E-01
K = 8 GAINS 4.90E-01	3. 238-01
K = 9 GAINS 4.87E-01	3. 20E-01
K = 10 GAINS 4.86E-01	3.21E-01
K = 11 GAINS 4.86E-01	3.21E-01
K = 12 GAINS 4.86E-01	3.215-01
K = 13 GAINS 4.86E-01	3.218-01
K = 14 GAINS 4.86E-01	3.21E-01
K = 15 GAINS 4.86 E-01	3.212-01
K = 16 GAINS 4.36E-01	3.216-01
K = 17 GAINS 4.86 E-01	3.21E-01
K = 18 GAINS 4.86E-01	3.218-01

K = 19 GAINS 4.86E-01 3.21E-01

K = 20 GAINS 4.86E-01 3.21E-01

THIS CONCLUDES THE DISCRETE KALMAN FILTER PROGRAM (KALMAN).

ANALYSIS IS COMPLETE. DO YOJ WANT TO RUN LINCON AGAIN? LINCON IS NOW TERMINALED.

The computer results are presented in Appendix 3.

# E. STATE VARIABLE FEEDBACK PROGRAM (STVAR) Given the linear time-invariant system

$$\underline{\dot{x}}(t) = \underline{\lambda}\underline{x}(t) + \underline{b}u(t) \qquad (4. E-1)$$

$$u(t) = K[r(t) - \frac{T}{k}x(t)]$$
 (4. E-2)

$$Y(t) = \mathbf{c}^{\mathsf{T}} \mathbf{x}(t) \tag{4. E.3}$$

the following can be performed by STVAR

- (1) calculation of the plant tranfer function, Y(s)/U(s)
- (2) by defining a ficticious  $\underline{c}$  vector the internal transfer function can be calculated,  $X_i(s)/U(s)$ ; for example, if  $X_3(s)/U(s)$  is desired the  $\underline{c}$  matrix is selected with  $c_3 = 1$  and all other  $\underline{c}$  elements equal to zero; or if  $X_1(s)/X_3(s)$  is desired, calculate  $X_1(s)/U(s)$  and  $X_3(s)/U(s)$  and divide the two
- (3) calculation of the closed-loop transfer function, Y(s)/R(s)
- (4) calculation of the feedback transfer function,  $H_{ee}$  (5)
- (5) for a desired closed-loop transfer function, the controller gain and feedback coefficients in addition to  $H_{eq}$  (s) can be calculated: the

feedforward gain is selected so that a zero steady state error results from a step input; the designer who wishes other conditions must rescale the gain and feedback coefficients appropriately; for example, if it is desired to have K = 1.0 but it is calculated as K = 2.0 with feedback coefficients of k = 0.5, k = 0.0 and k = 1.5, the procedure to modify the results would be

$$\underline{\mathbf{k}} = \frac{2.0}{1.0} - [0.5 \quad 0.0 \quad 1.5]$$

$$= [1.0 0.0 3.0]$$

All of the information necessary for the user to solve state variable feedback problems is presented in this section. However, the theoretical concepts involved in the development of the computer codes are fully described by Melsa [7].

The basic input contains the problem identification, matrices  $\underline{\mathbf{A}}$  and  $\underline{\mathbf{b}}$  and the order of the plant, n. These four inputs are required regardless of what open- or closed-loop calculations are to be made.

At this point STVAR verifies the controllability of the system. Three controllability conditions are possible

(1) complete controllability

- (2) numerically uncontrollable and
  - (3) uncontrollability

Contollability arises when the controllability matrix

 $\underline{\mathbf{E}} = \left[ \underline{\mathbf{b}} \ \underline{\mathbf{A}} \underline{\mathbf{b}} \ \underline{\mathbf{A}}^2 \underline{\mathbf{b}} \ \dots \ \underline{\mathbf{A}}^{n-1} \underline{\mathbf{b}} \right] \tag{4.2-4}$ 

is nonsingular, i.e., let # 3. Even if the matrix is nonsingular problems may arise if it is difficult to invert. To check this STVAR multiplies the matrix by its calculated inverse. The result should be the identity matrix. The actual matrix product is compared with the identity matrix to provide a measure of uncontrollability. If the maximum value of deviation is not negligible, the plant is identified as numerically uncontrollable. A deviation larger than 10-3 to 10-5 has been found to indicate difficulty by Melsa [ 1].

User beware: If the system is identified as being
uncontrollable, all open- and closed-loop calculations are
still performed!

There are three possible closed-loop options available, one for analysis purposes only and the other two for design purposes. After choosing the analysis option, and supplying STVAR with the feedforward gain K and feedback coefficient

matrix **k**, the program calculates the closed-loop characteristic polynomial and the numerator of the equivalent feedback transfer function.

The two design options are used to calculate the controller gain and the feedback coefficients necessary for a desired closed-loop characteristic polynomial. The polynomial may be entered in either polynomial form or factored form.

## 1. Terminal Session Example

This section contains a third order system for analysis by STVAR as presented by Melsa [ 1]. The state variable representation of the plant is given by

$$\dot{\mathbf{x}}(+) = \begin{bmatrix} -1.0 & 1.0 & 0.0 \\ 0.0 & 0.3 & 1.0 \\ 0.0 & -3.0 & 0.0 \end{bmatrix} \mathbf{x}(-) + \begin{bmatrix} 0.0 \\ 0.0 \\ 1.0 \end{bmatrix} \mathbf{u}(-)$$

$$Y(t) = [1.0 1.0 0.0] x(t)$$

For the open-loop case, it is desired to find the internal transfer function X (s)/U(s) and the plant transfer function Y(s)/U(s). A finition matrix, necessary to find the internal transfer function, is then

$$c = [0.0 0.0 1.0]$$

In addition, find the values of the feedforward gain and the feedback coefficients required to give a closed-loop transfer function of

Y(s) = 2(s+2) $R(s) = s^3+4s^2+6s+4$ 

lincon

EXECUTION BEGINS...
LINCON CONSISTS OF THE FOLLOWING SUBPROGRAMS:
BASIC MATRIX MANIPULATION - <BASMAT>
RATIONAL TIME RESPONSE - <RIRESP>
STATE VARIABLE FEEDBACK - <SIVAR>
CONTROL LABILITY AND OBSERVABILITY - <OBSCON>
LUENBERGER OBSERVER - <LUEN>
OPTIMAL CONTROL/KALMAN FILTERS - <RICATI>
DISCRETE TIME KALMAN FILTER - <KALMAN>
OPTIMAL CONTROL - <OPICON>
PARTIAL FRACTION EXPANSION - <PRFEXP>
ROOTS OF A POLYNOMIAL - <ROOTS>
TO USE ONE OF THE SJBPROGRAMS ENTER THE NAME BETWEEN THE SYMBOLS < >.
Stwat

STVAR DETERMINES INTERNAL TRANSFER FUNCTIONS, THE PLANT TRANSFER FUNCTION, THE CLOSED-LOOP TRANSFER FUNCTION, AND THE EQUIVALENT FEEDBACK TRANSFER FUNCTION. IN ADDITION, THE CONTROLLER GAIN AND THE FEEDBACK COEFFICIENTS NECESSARY TO ACHIEVE A SPECIFIED CLOSED-LOOP TRANSFER FUNCTION ARE CALCULATED.

FIRST ENTER THE PROBLEM IDENTIFICATION (\*NOT TO EXCEED 2) CHARACTERS\*). thesis example

NOW, ENTER THE ORDER OF THE SYSTEM (UP TO 8).

ENTER THE ELEMENTS OF THE PLANT MATRIX--A
4 (1,1) =

 $\frac{\lambda(1,2)}{1}$ 

 $0 \quad A(1,3) =$ 

A (2,1) =

A(2,2) =

1. A(2,3) =

 $0 \quad A(3,1) =$ 

```
-3. 1 (3,2) =
   \lambda (3,3) =
     THE A MATRIX (PLANT MATRIX)
-1.00E+00 1.00E+00 0.0
0.0 0.0 1.00E+00
0.0 -3.00E+00 0.0
    DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
   ENTER THE ELEMENTS OF THE CONTROL VECTOR--B. B(1) =
    B(2) =
0
    B(3) =
1.
     THE B MATRIX (CONTROL MATRIX)
          1.00E+00
    DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
     OPEN-LOOP CALCULATIONS
     DENOMINATOR CDEFFICIENTS - IN ASCENDING POWERS OF S 3.00E+00 3.00E+00 1.00E+00
     THE ROOTS ARE
    DO YOU HAVE A FICTICIOUS DUIPUT VECTOR TO ENTER?
y
    ENTER THE ELEMENTS OF THE FICTICIOUS DUTPUT VECTOR--C. C(1) =
0
    C(2) =
0
    C(3) =
1.
     THE C MATRIX (FICTICIOUS DIPUT VECTOR)
          1.00E+00
    DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
n
     NUMERATOR COEFFICIENTS - IN ASCENDING POWERS OF S
                      1.00E+00 1.00E+00
```

```
REAL PART IMAG. PART
-1.00E+00 0.0
0.0
       THE ROOTS ARE
      DO YOU HAVE ANOTHER FICTICIOUS MATRIX TO ENTER?
n
     ENTER THE ELEMENTS OF THE TRUE OUTPUT VECTOR--C. C(1) =
1.
      C(2) =
1.
     C(3) =
0
       THE C MATRIX (OUIPUT VECIOR)
               1.00E+00
      DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
n
       NUMERATOR COEFFICIENTS - IN ASCENDING POWERS OF S 2.00E+00 1.00E+00
      THE ROOTS ARE

REAL PART IMAG. PART

-2.00 + 00 0.0

THIS PROGRAM IS CAPABLE OF PERFORMING THREE TYPES
OF CLOSED-LOOP CALCULATIONS. ONE TYPE FOR THE ANALYSIS
MODE. THE OTHER THO FOR DESIGN.
      ENTER ONE OF THE FOLLOWING MODES:

1) A -- FOR THE ANALYSIS MODE

2) P -- FOR THE DESIGN MODE WITH THE UNFACTORED CLOSED-LOOP CHARACTERISTIC POLYNOMIAL

3) F -- FOR THE DESIGN MODE WITH THE FACTORED
                           CLOS ED-LOOP
                                                CHARACTERISTIC POLYNOMIAL
þ
       CLOSED-LOOP CALCILATIONS
      KEY = P *****
ENTER THE DESIRED CLOSED-LOOP CHARACTERISTIC POLYNOMIAL COEFFICIENTS IN ASCENDING POWERS OF S.
YOUR HIGHEST ORDER COEFFICIENT MUST BE A VALUE OF ONE.
      POLY (1) =
      POLY(2) =
      POLY(3) =
      POLY (4) =
1.
        THE NUMERATOR OF H-EQUIVALENT -
IN ASCENDING POWERS OF S
5.00E-01 1.50E+00 1.50E+00
```

THE ROOTS ARE

REAL PART IMAG. PART -5.00E-01 -2.89E-01 -5.00E-01 2.89E-01 THE FEEDBACK COEFFICIENTS 5.00E-01 0.0 1.50E+00

THE GAIN = 2.000000E+00

THE CLOSED-LOOP CHARACTERISTIC POLYNOMIAL - IN ASCENDING POWERS OF S 4.00E+00 6.00E+00 4.00E+00 1.00E+00

THE ROOTS ARE REAL PART IMAG. PART
-2.00E+00 0.0
-1.00E+00 -1.00E+00
-1.00E+00 1.00E+00

MAXIMUM NORMALIZED ERROR = 0.0

DO YOU WANT TO RUN ANOTHER MODE IN STVAR?

THIS CONCLUDES THE STATE VARIABLE FEEDBACK PROGRAM (STVAR) .

ANALYSIS IS COMPLETE. DO YOJ WANT TO RUN LINCON AGAIN?
LINCON IS NOW TERMINATED.

The computer results for this problem is presented in Appendix H. The open-loop portion indicates that the system is controllable, since there is no indication of uncontrollability, with the desired internal transfer function of

$$\frac{X_3}{3} \frac{(s)}{(s)} = \frac{s^2 + s}{s^3 + s^2 + 3s + 3}$$

and a plant transfer function of

$$Y(s) = s+2$$
  
 $U(s) = s^3 + s^2 + 3s + 3$ 

The closed-loop portion, using the polynomial design mode, shows that the feedforward gain and feedback

coefficient matrix required to yield the desired closed-loop transfer function, Y(s)/R(s), are

$$K = 2.0$$

an ā

$$\underline{\mathbf{k}} = \begin{bmatrix} 0.5 \\ 0.0 \\ 1.5 \end{bmatrix}$$

#### F. LUENBERGER OBSERVER PROGRAM (LUEN)

When a particular closed-loop transfer function is desired and some of the states are inaccessible, LUEN can be used to design a Luanberger Observer. For example, if q measurements are state variables, an observer of reduced dimensions can be designed to estimate only those states which are not measured. The state estimates generated by an observer can be used as input information to a controller [8]. The block diagram presented in Fig. 4-1 represents the general form of the system when a compensator is placed in the feedback path.

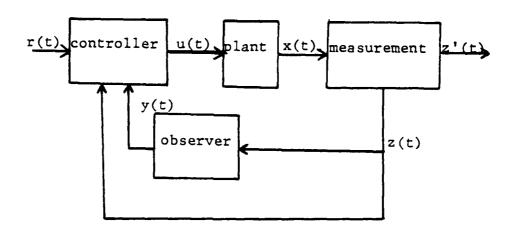


Fig. 4.1. Luenberger Observer Block Diagram

The plant is characterized by the state and measurement equations

$$\dot{\mathbf{x}}(t) = \mathbf{A}\dot{\mathbf{x}}(t) + \mathbf{B}\mathbf{u}(t) \tag{4. F-1}$$

$$\underline{z}(t) = \underline{C}\underline{x}(t) \tag{4.7-2}$$

$$z^{*}(t) = \underline{c} \underline{x}(t) \tag{4. F-3}$$

An observer can be designed that generates an estimate  $\underline{y}$  which converges to the state  $\underline{x}$  as time becomes large [ 9]. A linear controller is designed as

$$\underline{\mathbf{u}}(t) = K[\underline{\mathbf{r}}(t) - \underline{\mathbf{k}}^{\mathsf{T}}\underline{\mathbf{x}}(t)] \tag{4. F-4}$$

in which all states can be measured. Replacing the true state with the estimate yields

$$\underline{\mathbf{u}}(t) = \mathbf{K}[\underline{\mathbf{r}}(t) - \underline{\mathbf{k}}^{\mathsf{T}}\underline{\mathbf{v}}(t)] \tag{4. F-5}$$

where

$$\underline{k}^{\mathsf{T}} \underline{\mathbf{y}}(t) = \underline{h}^{\mathsf{T}} \underline{\mathbf{y}}(t) + \underline{\mathbf{q}}^{\mathsf{T}} \underline{\mathbf{z}}(t) \tag{4.F-6}$$

As time increases  $\underline{k}^T\underline{y}(t)$  will approach  $\underline{k}^T\underline{x}(t)$ . Since the reduced-order observer estimates only the inaccessible states, the control law contains measured states where available and observer estimates for the other states.

For ease in using LUEN, the following are defined:

- x(t) = n-element column state vector
- $\underline{u}(t) = plant in put$
- $\underline{z}(t) = q$ -vector of system measurements
- z'(t) = output variable to be controlled
  - $\underline{\mathbf{A}}$  = plant matrix (n x n)
- $\underline{b}$  = distribution matrix (q x m)
  - [assumptions: (1)  $q \le n$  and (2) C has rank q]
- C = output matrix
- r(t) = forcing function
- K = feedforward or controller gain
- $\underline{\mathbf{k}}$  = feedback coefficient matrix(n x m)
- b = observer feedback coefficient matrix
- g = output feedback coefficient matrix

an d

 $\mathbf{Y}(t)$  = estimated state vector of  $\mathbf{x}(t)$ .

From theory, the observer is described by

$$Y(t) = ZY(t) + G_1Z(t) + G_2U(t)$$
 (4. P-7)

where

<u>F</u> = observer eigenvalue matrix

and

 $G_1$  and  $G_2$  = observer gain matrices.

 $Z^*(s)/R(s)$  can be solved by the following procedure:

- (1) select Z'(s)/R(s) and solve for K and k by using the STVAR program
- (2) use the OBSCON program to calculate the observability index; the observer is designed to have a dimension greater than or equal to the observability index minus one.
- (3) select the P eigenvalues: these should not equal those of A, previously calculated by STVAR.
- (4) use LUEN to calculate G, , G, h and q.
- 1. Terminal Session Example

The example presented hare is taken from Desjardins, pp.147 [ 2]. Commands entered by the user are in lower

case. However, due to the special nature of the example, comments have been added to the terminal session for clarity. Given the fourth order plant

$$\dot{\mathbf{x}}(t) = \begin{bmatrix} 0.0 & 1.0 & 0.0 & 0.0 \\ 0.0 & 3.0 & 1.3 & 0.0 \\ 3.0 & -15. & -23. & -9.0 \end{bmatrix} \mathbf{x}(t) + \begin{bmatrix} 0.0 \\ 0.0 \\ 0.0 \\ 1.0 \end{bmatrix} \mathbf{u}(t)$$

with

$$z^*(t) = [10. 20. 0.0 0.0] \underline{x}(t)$$

Because  $x_1$  and  $x_2$  are the only measureable states

$$\underline{\mathbf{y}}$$
 (t) = \[ \begin{pmatrix} 1.0 & 0.0 & 0.0 & 0.0 \\ 2.0 & 1.0 & 0.0 & 0.0 \end{pmatrix} \, \bar{\textbf{x}} \text{(t)} \]

The closed-loop transfer function to be achieved is chosen to be

$$\frac{Z'(s)}{R(s)} = \frac{1}{s^4 + 5s^3 + 17s^2 + 28s + 20}$$

lincon

EXECUTION BEGINS...
LINCON CONSISTS OF THE FOLLOWING SUBPROGRAMS:
BASIC MATRIX MANIPULATION - <BASMAT>
RATIONAL TIME RESPONSE - <RTRESP>
STATE VARIABLE FEEDBACK - <STVAR>
CONTROL LABILITY AND OBSERVABILITY - <OBSCON>
LUENBERGER OBSERVER - <LUEN>
OPTIMAL CONTROL/KALMAN FILTERS - <RICATI>
DISCRETE TIME KALMAN FILTER - <KALMAN>
OPTIMAL CONTROL - <OPTCON>
PARTIAL FRACTION EXPANSION - <PRFEXP>

ROOTS OF A POLYNOMIAL - <ROOTS>
TO USE ONE OF THE SUBPROGRAMS ENTER THE NAME BETWEEN THE SYMBOLS < >.
stvai

STVAR DETERMINES INTERNAL TRANSPER FUNCTIONS, THE PLANT TRANSFER FUNCTION, THE CLOSED-LOOP TRANSFER FUNCTION, AND THE EQUIVALENT FEEDBACK TRANSFER FUNCTION. IN ADDITION, THE CONTROLLER GAIN AND THE FEEDBACK COEFFICIENTS NECESSARY TO ACHIEVE A SPECIFIED CLOSED-LOOP TRANSFER FUNCTION ARE CALCULATED.

FIRST ENTER THE PROBLEM IDENTIFICATION (\*NOT TO EXCEED 23 CHARACTERS\*).
THESIS EXAMPLE

NOW, ENTER THE ORDER OF THE SYSTEM (UP TO 8).

ENTER THE ELEMENTS OF THE PLANT MATRIX -- A

- A(1,1)=
- A (1, 2) =
- A(1,3) =
- A (1, 4) =
- $0^{A(2,1)}=$
- $_{0} \quad \lambda \left( 2,2\right) =$
- A(2,3) =
- A(2,4) =
- A(3, 1) =
- $0 \quad A(3,2) =$
- A(3,3) =
- A(3,4) =
- A (4, 1) =
- -15.<sup>A (4, 2) =</sup>
- -23. A (4, 3) =
- -9. A (4, 4) =

```
THE A MATRIX (PLANT MATRIX)
                   1.30E+00 0.3
0.0 1.30E+00 0.3
0.3 0.30E+01 -9.03E+00
   DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
   ENTER THE ELEMENTS OF THE CONTROL VECTOR -- B.
   B (1) =
   B(2) =
   B(3) =
   B (4) =
    THE B MATRIX (CONTROL MATRIX)
         8:8
         0.0
1.00E+00
   DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
******************
    OPEN-LOOP CALCULATIONS
    DENOMINATOR COEFFICIENTS - IN ASCENDING POWERS OF S
                    1.50E+01 2.30E+01 9.00E+00 1.00E+00
         0.0
                           REAL PART IMAG. PART
-3.00E+00 0.0
-5.00E+00 0.0
0.0 0.0
    THE ROOTS ARE
   DO YOU HAVE A FICTICIOUS DUTPUT VECTOR TO ENTER?
   ENTER THE ELEMENTS OF THE TRUE OUTPUT VECTOR -- C.
20.<sup>C(1)=</sup>
10.<sup>C</sup>(2)=
   C(3) =
   C(4) =
    THE C MATRIX (OUTPUT VECIOR)
2.00E+01
1.00E+01
0.0
0.0
```

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

NUMERATOR COEFFICIENTS - IN ASCENDING POWERS OF S 2.00E+01 1.00E+01

THE ROOTS ARE

REAL PART IMAG. PART -2.00E+00 0.0

THIS PROGRAM IS CAPABLE OF PERFORMING THREE TYPES OF CLOSED-LOOP CALCULATIONS. ONE TYPE FOR THE ANALYSIS MODE. THE OTHER INO FOR DESIGN.

- ENTER ONE OF THE FOLLOWING MODES:

  1) A -- FOR THE ANALYSIS MODE

  2) P -- FOR THE DESIGN MODE WITH THE UNFACTORED CLOSED-LOOP CHARACTERISTIC POLYNOMIAL

  3) F -- FOR THE DESIGN MODE WITH THE FACTORED CLOSED-LOOP CHARACTERISTIC POLYNOMIAL

þ

n

CLOSED-LOOP CALCILATIONS

KEY = P \*\*\*\*\*
ENTER THE DESIRED CLOSED-LOOP CHARACTERISTIC POLYNOMIAL COEFFICIENTS IN ASCENDING POWERS OF S.

YOUR HIGHEST ORDER COEFFICIENT MUST BE A VALUE OF ONE. POLY(1) = 20.

POLY(2) =28.

POLY(3) =17.

POLY(4) =6 -

POLY(5) =

THE NUMERATOR OF H-E) UIVALENT - IN ASCENDING POWERS OF S

2.00E+01 1.30E+01 -5.30E+00 -3.03E+00

REAL PART IMAG. PART -1.04E+00 0.0 3.21E+00 0.0 THE ROOTS ARE

THE FEEDBACK COEFFICIENTS

2.00E+01 1.30E+01 -6.00E+00 -3.00E+00

THE GAIN = 1.000000E+33

THE CLOSED-LOOP CHARACTERISTIC POLYNOMIAL - IN ASCENDING POWERS OF S

2.00E+01 2.30E+01 1.70E+01 6.00E+00 1.00E+00

REAL PART IMAG PART
-1.00E+00 -2.00E+00
-2.00E+00 0.0
-2.00E+00 0.0 THE ROOTS ARE

MAXIMUM NORMALIZED ERROR =

DO YOU WANT TO RUN ANOTHER MODE IN STVAR?

THIS CONCLUDES THE STATE VARIABLE PEEDBACK PROGRAM (STVAR).

COMMENT: results indicate that the system is controllable, that the eigenvalues of A are -3.0, -5.0, -1.0 and 0.0, that the values of k are 20.0, 13.0, -6.0 and -3.0 and that K is equal to unity; the observability index is next needed to design the observer

ANALYSIS IS COMPLETE. DO YOJ WANT TO RUN LINCON AGAIN? y

LINCON CONSISTS OF THE FOLLOWING SUBPROGRAMS:

BASIC MATRIX MANIPULATION - <BASMAT>

RATIONAL TIME RESPONSE - <RIRESP>

STATE VARIABLE FEEDBACK - <SIVAR>
CONTROL LABILITY AND OBSERVABILITY - <OBSCON>
LUENBERGER OBSERVER - <LUEN>
OPTIMAL CONTROL/KALMAN FILIERS - <RICATI>
DISCRETE TIME KALMAN FILIER + <KALMAN>
OPTIMAL CONTROL - <OPICON>
PARTIAL FRACTION EXPANSION - <PRFEXP>
ROOTS OF A POLYNOMIAL - <ROOIS>
TO USE ONE OF THE SUBPROGRAMS ENTER THE NAME BETWEEN THE SUBBOLS <>CODSCOR obscon

OBSCON DETERMINES THE OBSERVABILITY INDEX AND CONTROL LABILITY OF A SYSTEM.

FIRST ENTER THE PROBLEM IDENTIFICATION (\*NOT TO EXCEED 23 CHARACTERS\*). thesis example

NOW, ENTER THE ORDER OF THE SYSTEM (UP TO 8).

ENTER THE ELEMENIS OF THE PLANT MATRIX -- A.

A(1,1) =0

 $\lambda(1,2) =$ 1.

A(1,3) =

A(1,4) =

A(2, 1) =

```
A(2,2) =
   A(2,3) =
   A(2,4) =
   A(3, 1) =
   A(3, 2) =
   A(3,3) =
   A(3,4) =
   A(4,1) =
-15<sup>A</sup>(4,2) =
-23. A (4,3) =
-9. A (4,4) =
     THE A MATRIX (PLANT MATRIX)
                        ]:00E+00
    DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
    ENTER THE NUMBER OF COLUMNS OF THE B MAIRIX.
   ENTER THE ELEMENTS OF THE B MATRIX. B (1,1) =
0
    B(2, 1) =
0
    B(3, 1) =
   B(4, 1) =
     THE B MATRIX
 0.5
1.00E+00
```

DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

```
ENTER THE NUMBER OF OUTPUTS.
2
      ENTER THE ELEMENIS OF THE C MATRIX. C(1,1) =
1.
      C(1,2) =
0
      C(1,3) =
      C(1,4) =
      C(2,1) =
0
      C(2,2) =
1.
      C(2,3) =
0
      C(2,4) =
1.00 THE C MATRIX
                                            Q.Ç
                       1. 0 OE+00
                                           0.0
       DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
         OBSERVABILITY INDEX =
THE SYSTEM (A.B) IS CONTROLLABLE
      DO YOU WANT TO RUN OBSCON AGAIN?
       THIS CONCLUDES THE OBSERVABILITY INDEX AND CONTROLIABILITY PROGRAM (OBSCON)
               COMMENT: results indicate an observability index of 3; this permits the design of a second order observer; eigenvalues of -3.5 and -4.0 are chosen (not equal to those of the plant) as observer eigenvalues
       ANALYSIS IS COMPLETE. DO YOU WANT TO RUN LINCON AGAIN?
LINCON CONSISTS OF THE FOLLOWING SUBPROGRAMS:

BASIC MATRIX MANIPULAITON - <BASMAT>
RATIONAL TIME RESPONSE - <RIRESP>
STAFE VARIABLE FEEDBACK - <SIVAR>
CONTROLIABILITY AND OBSERVABILITY - <OBSCON>
LUENBERGER OBSERVER - <LUEN>
OPTIMAL CONTROL/KA LMAN FILTERS - <RICATI>
DISCRETE TIME KALMAN FILTERS - <KALMAN>
OPTIMAL CONTROL / COPICON>
PARTIAL PRACTION EXPANSION - <PREXP>
ROOTS OF A POLYNOMIAL - <ROOTS>
TO USE ONE OF THE SUBPROGRAMS ENTER THE NAME BETWEEN THE SYMBOLS < >.
luen
```

LUEN IS USED TO DESIGN LUENBERGER OBSERVERS TO ACHIEVE A GIVEN CLOSED-LOOP TRANSFER FUNCTION WHEN SOME STATE VARIABLES ARE INACCESSIBLE.

FIRST ENTER THE PROBLEM IDENTIFICATION (NOT TO EXCEED 20 CHARACTERS). thesis example

- ENTER THE ORDER OF THE SYSTEM (UP TO 8).
- ENTER THE NUMBER OF MEASUREMENTS (UP TO 8).
- ENTER THE ORDER OF THE OBSERVER (UP TO 8).

ENTER THE ELEMENTS OF THE PLANT MATRIX -- A.

- $0 \quad A(1,1) =$
- 1.  $\lambda(1,2) =$
- $0 \quad A(1,3) =$
- $0 \quad A(1,4) =$
- $0 \quad A(2,1) =$
- A(2,2) =
- 1. A(2,3) =
- $0 \quad A(2,4) =$
- a(3,1) =
- A(3,2) =
- $_{0} \quad A(3,3) =$
- A(3,4) =
- A(4,1) =
- -15<sup>A</sup>(4,2) =
- -23.4(4,3) =
- -9. A (4,4) =

```
THE A MATRIX (PLANT MATRIX)
0.0 1.00E+00 3.1
0.0 0.3 1.30E+00
0.0 0.3 3.30E+01
                      0.3 1.30E+00 U.U
0.3 3.3 1.00E+00
-1.50E+01 -2.30E+01 -9.00E+00
    DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
n
    ENTER THE ELEMENTS OF THE DISTRIBUTION MATRIX -- B.
    B(1) =
0
    B(2) =
    B(3) =
0
    B(4) =
1.
     THE B MATRIX (DISTRIBUTION MATRIX)
0.0
0.0
0.0
          1.00E+00
    DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
n
    ENTER THE ELEMENIS OF THE OUTPUT MATRIX -- C.
    C(1,1) =
1.
    C(1,2) =
    C(1,3) =
a
    C(1,4) =
0
    C(2, 1) =
0
    C(2,2) =
    C(2,3) =
0
    C(2,4) =
0
     THE C MATRIX (OJIPUT MATRIX)
    J. 0 1.00E+00 J.J 0.0
DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
\mathbf{r}
    ENTER THE DESIRED FEEDBACK COEFFICIENTS.
FDBK COEFF(1) = 20.
```

```
FDBK COEFF (2) = 13.
```

PDBK COEFF (3) =

-3. FDBK COEFF (4) =

THE DESIRED FEEDBACK COEFFICIENTS

2.00E+01 1.30E+01 -6.00E+00 -3.00E+00 DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?

n

THE OBSERVER EIGENVALUES (F MATRIX) CAN BE SUPPLIED EITHER IN THE FORM OF A CHARACTERISTIC POLYNOMIAL OR IN THE ROOTS OF THAT POLYNOMIAL.

ENTER EITHER A P FOR POLYNOMIAL COEFFICIENT FORM OR AN F FOR FACIORED ROOF FORM.

ENTER THE REAL PART OF THE ROOT.

ENTER THE MAGNITUDE OF THE IMAGINARY ROOT.

ENTER THE REAL PART OF THE ROOT.

ENTER THE MAGNITUDE OF THE IMAGINARY ROOT. 0

OBSERVER EIGENVALUES

REAL PART IMAG. PART -3.502+00 0.0 -4.00E+00 0.0

THE OBSERVER CHARACTERISTIC POLYNOMIAL COEFFICIENTS IN ASCENDING POWERS OF S 1.40E+01 7.50E+00 1.33E+00

EIGENVALUE MAIRIX)

THE F MATRIX (OBS ERVER -7.50E+00 1.30E+00 -1.40E+01 0.3

THE 31 MATRIX (OBSERVER GAIN MATRIX)
9.55E+01 2.92E+01
0.0

THE 32 MATRIX (OBSERVER GAIN MATRIX)
-3.00E+00
-1.50E+00

OUTPUT FEEDBACK COEFFICIENTS 2.00E+01 8.50E+00

COMPENSATOR FEEDBACK COEFFICIENTS
1.00E+00 0.3

THIS CONCLUDES THE LUENBERGER OBSERVER DESIGN PROGRAM.

DO YOU WANT TO RUN THE PROGRAM AGAIN?

ANALYSIS IS COMPLETE. DO YOJ WANT TO RUN LINCON AGAIN?

LINCON IS NOW PERMINAPED.

The results of STVAR, DBSCON and LUEN are shown in Appendix I. From these the the observer is given as

$$y(t) = \begin{bmatrix} -7.5 & 1.0 \end{bmatrix} \begin{bmatrix} y_3(t) \end{bmatrix} + \begin{bmatrix} 85.5 & 29.25 \end{bmatrix} \begin{bmatrix} x_1(t) \end{bmatrix} + \begin{bmatrix} -3.0 \end{bmatrix} u(t)$$

$$= \begin{bmatrix} -14.0 & 0.0 \end{bmatrix} \begin{bmatrix} y_+(t) \end{bmatrix} + \begin{bmatrix} 0.0 & 0.0 \end{bmatrix} \begin{bmatrix} x_2(t) \end{bmatrix} + \begin{bmatrix} -1.5 \end{bmatrix} u(t)$$

and

$$u(t) = 1.0 \{ t(t) - [20.0 3.5] \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} - [1.0 0.0] \begin{bmatrix} y_3(t) \\ y_4(t) \end{bmatrix} \}$$

#### G. OPTIMAL CONTROL PROGRAM (OPTCOM)

Given the linear, time-invariant system represented as

$$\dot{\mathbf{x}}(t) = \underline{\mathbf{A}}\mathbf{x}(t) + \underline{\mathbf{B}}\mathbf{u}(t) \tag{4. G-1}$$

OPTCON will minimize the cost function

$$J(N) = 1/2\underline{x} (N) Q\underline{x}(N) + 1/2 \sum_{k=0}^{M-1} [\underline{x}(k) 2\underline{x}(k) + Ru^{2}(k)]$$
 (4. G-2)

where

 $\underline{\mathbf{x}}$  = state vector

Q = measurement noise covariance matrix (n x n)

N = number of time intervals over which the sum is made

R = random input (a scalar)

 $\underline{\mathbf{A}}$  = plant matrix (n x n)

B = distribution matrix (n x 1)

anā

u(t) = control (a scalar).

The output of the program is the feedback gain matrix which, when multiplied by the state vector, yields a scalar control. The following recursive equations were derived using dynamic programming, starting at the terminal time and working backwards.

$$\underline{\mathbf{p}}(\mathbf{k}) = \underbrace{\psi}^{\mathsf{T}}(\mathbf{k})\underline{\mathbf{p}}(\mathbf{k}-1)\underbrace{\psi}(\mathbf{k}) + \underline{\mathbf{q}} + \mathbf{R}\underline{\mathbf{A}}(\mathbf{k})\underline{\mathbf{A}}^{\mathsf{T}}(\mathbf{k}), \ \underline{\mathbf{p}}(\mathbf{0}) = 0 \qquad (4.3-3)$$

$$\underline{\psi}_{(k)} = \underline{\Phi} + \underline{\Delta}\underline{A}^{\mathsf{T}}(k), \ \underline{\psi}_{(0)} = 0 \tag{4.6-4}$$

$$\underline{\underline{\mathbf{A}}}^{\mathsf{T}}(k) = -[\underline{\Delta}\underline{P}(k-1)\underline{\underline{\Phi}}]/[\underline{\Delta}^{\mathsf{T}}\underline{P}(k-1)\underline{\Delta} + R], \ \underline{\underline{\mathbf{A}}}^{\mathsf{T}}(0) = 0 \qquad (4.6-4)$$

For simplicity in programming, the following terms are defined:

terminal = 
$$1/2\underline{x}^{T}(N)\underline{Q}\underline{x}(N)$$
  
trajectory =  $1/2\sum_{k=0}^{N-1}\underline{x}(k)\underline{Q}\underline{x}(k)$   
fuel =  $1/2\sum_{k=0}^{N-1}Ru^{2}(k)$ 

### 1. Terminal Session Example

Given the system and parameters described below find the discrete steady state gains for a sample of 0.1.

$$\dot{\mathbf{x}}(t) = \begin{bmatrix} 0.0 & 1.0 \\ 1.0 & 0.0 \end{bmatrix} \mathbf{x}(t) + \begin{bmatrix} 0.0 \\ 1.0 \end{bmatrix} \mathbf{u}(t)$$

$$Q = \begin{bmatrix} 1.0 & 1.0 \\ 1.0 & 1.0 \end{bmatrix}$$

R = 1.0

In addition, run the program for a time interval of 40.

lincon

EXECUTION BEGINS ...

```
LINCON CONSISTS OF THE FOLLOWING SUBPROGRAMS:

BASIC MATRIX MANIPULATION - (BASMAT)

RATIONAL TIME RESPONSE - (RIRESP)

STATE VARIABLE FEEDBACK - (SIVAR)

CONTROLIABILITY AND OBSERVABILITY - (OBSCON)

LUENBERGER OBSERVER - (LUEN)

OPTIMAL CONTROL/KALMAN FILTERS - (RICATI)

DISCRETE TIME KALMAN FILTER - (KALMAN)

OPTIMAL CONTROL - (OPTION)

PARTIAL FRACTION EXPANSION - (PRFEXP)

ROOTS OF A POLYNOMIAL - (ROOTS)

TO USF ONE OF THE SUBPROGRAMS ENTER THE NAME BETWEEN THE SYMBOLS < >.
  optcon
         OPTCON MINIMIZES THE THE FOLLOWING COST FUNCTION:
                 J(N) = MIN(SUM(X(N)T*Q*X(N)+UT(N-1)*R*U(N-1)))
         THE OUTPUT OF THE PROGRAM IS THE FEEDBACK GAIN MATRIX, A TRANSFOSE, WHICH WHEN MULTIPLIED BY THE STATE VECTOR YIELDS A SCALAR CONTROL.
THE FOLIOWING RECURSIVE EQUATIONS WERE DERIVED USING DYNAMIC PROGRAMMING, STARTING AT THE TERMINAL TIME AND WORKING BACKWARDS:
                          AT (K) = -(DELT*P(K-1)*PHI/(DELT*P(K-1)*DEL+R)
                                                                                                                 AT(0) = 0
                                                                                                                                 PSI(0) =
                 (2)
                                  PSI(K) = PHI + DEL * AT(K)
0
                 (3) P(K) = PSIT(K) * P(K-1) * PSI(K) + Q+R*A(K) * AT(K) P(0) = 0
         FIPST ENTER THE PROBLEM IDENTIFICATION (*NOT TO EXCEED 2) CHARACTERS*). thesis example
          ENTER THE NUMBER OF TIME INTERVALS (NSTAGE) OVER WHICH THE SUM IS TO BE MADE.

*** NSTAGE MUST BE ENTERED IN 13 FORMAT ***

*** (I.E., PIGHT JUSTIFY TO THREE DIGITS ***
  040
         ENTER THE ORDER OF THE SYSTEM (UP TO 8).
*** ENTER IN I1 FORMAT ***
  2
         ENTER THE ELEMENTS OF THE O MATRIX.

*** ALL MATRICES ARE ENTERED IN F-FORMAT ***

*** I.E., PUT A DECIMAL POINT AFTER YOUR NUM
                                                                                          *** REMNUM RUCY RETTA
         Q(1, 1) =
          Q(1,2) =
  0
          Q(2,1) =
  0
          Q(2,2) =
          THE Q MAIRIX
1.30E+00 0.0
0.0 2.0
          0.0 2.00E+33
DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
  n
          ENTER THE VALUE OF THE SCALAR R
   1.
```

```
ENTER THE SAMPLE INTERVAL--DT.
*** ENTER DT IN F-FORMAT ***
. 1
AT THIS POINT YOU MUST CHOOSE ONE OF THE FOLLOWING OPTIONS:
OPTION A: ENTER THE NUMBER O IF

(1) R IS FINITE, COST=TERMINAL+TRAJECTORY+FUEL,
OR IF
(2) R IS ZERO, COST=TERMINAL+TRAJECTORY+ O
           OPTION B: ENTER THE NUMBER 1 IF
(1) R IS FINITE, COST=TERMINAL+ 3 + FUEL,
OR IF
(2) R IS ZERO, COST=TERMINAL + 0 + 0
     IF YOU WANT TO READ IN THE A AND B MAIRICES, BUT NOT THE PHI AND DEL MATRICES, ENTER A O
      HOWEVER, IF YOU WANT TO ENTER THE PHI AND DEL MATRICES, BUT NOT THE A AND B MATRICES, ENTER A 1.
     ENTER THE ELEMENTS OF THE PLANT MATRIX--A. A (1, 1) =
0
      A(1,2) =
1.
     A(2, 1) =
      A(2, 2) =
      THE A MAIRIX (PLANT MATRIX)
0.0 1.00E+00
1.00E+00
      DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
     ENTER THE ELEMENIS OF THE DISTRIBUTION MATRIX--B. B (1, 1) \approx
0
      B(2, 1) =
      THE B MATRIX (DISTRIBUTION MATRIX)
      1.00E+00
      DO YOU WANT TO CHANGE ANY ELEMENT OF THE MATRIX?
       THE PHI MATRIX
1.005003E+00
1.001667E-01
                                         1.301657E-01
1.005003E+00
       THE DEL MATRIX
5.004164E-03
1.001667E-01
     MINIMIZATION OVER ALL STAGES

J(1) = AT(J) *X(J)

AT(1) = -2.46E-J2

AT(2) = -1.98E-J1
```

U(2) = AT(JL\*X(J) AT(1) = -7.72E-J2 AT(2) = -3.91E-J1U(3) = AT(J) \*X(J) AT(1) = -1.55E-01 AT(2) = -5.79E-01U(4) = AT(J) \*X(J) AT(1) = -2.54E-01 AT(2) = -7.58E-01U(5) = AT(J) \* X(J) AT(1) = -3.70E-01 AT(2) = -9.29E-01U(6) = AT(J) \*X(J) AT(1) = -4.98E-J1 AT(2) = -1.09E+JJ $\begin{array}{l} U(7) = AT(J) * X(J) \\ AT(1) = -6.34E-J1 \\ AT(2) = -1.24E+JJ \end{array}$ U(8) = AT(J) \* X(J) AT(1) = -7.74E-01 AT(2) = -1.38E+00U(9) = AT(J) \* X(J) AT(1) = -9.12E-J1 AT(2) = -1.50E+OJU(10) = AT(J)\*X(J) AT(1) = -1.05E+30 AT(2) = -1.61E+33U(11) = AT(J) \*X(J) AT(1) = -1.18E+30 AT(2) = -1.72E+30U(12) = AT(J) \*X(J) AT(1) = -1.30E+33 AT(2) = -1.81E+33U(13) = AT(J) \*X(J) AT(1) = -1.41E+J3 AT(2) = -1.89E+33U(14) = AT(J) \* X(J) AT(1) = -1.51E+00 AT(2) = -1.96E+00U(15) = AT(J) \*X(J) AT(1) = -1.60E+33 AT(2) = -2.02E+03U(16) = AT(J)\*X(J) AT(1) = -1.68E+30 AT(2) = -2.07E+30U(17) = AT(J) \*X(J) AT(1) = -1.76E+JJ AT(2) = -2.12E+JJU(18) = AT(J) \* X(J) AT(1) = -1.82E+33 AT(2) = -2.16E+33

 $\begin{array}{ccc} & \text{U(36)} &=& \text{AT(J)} * \text{X(J)} \\ & \text{AT(2)} &=& -2 \cdot 21 \text{E+33} \\ & & -2 \cdot 39 \text{E+33} \end{array}$ 

U(37) = AT(J) \*X(J) AT(1) = -2.22E+03 AT(2) = -2.39E+03

U(38) = AT(J) \*X(J) AT(1) = -2.22E+J3 AT(2) = -2.40E+J3

U(39) = AT(J) \*X(J) AT(1) = -2.22E+33 AT(2) = -2.40E+33

U(40) = AT(J) \* X(J) AT(1) = -2.23E+33

THIS CONCLUDES THE OPTIMAL CONTROL PROGRAM (OPTCON). DO YOU WANT TO RUN THE PROGRAM AGAIN? n

ANALYSIS IS COMPLETE. DO YOU WANT TO RUN LINCON AGAIN? LINCON IS NOW TERMINATED.

The results, shown in Appendix J, indicate that steady state gains are achieved at about -2.23 and -2.4.

#### V. CONCLUSIONS AND RECOMMENDATIONS

#### A. CONCLUSIONS

Although LINCON was written primarily as a teaching/learning tool, it can still be quite useful to the practicing engineer for design and analysis problems. It was written in modular form so that it could be easily modified by the addition of subroutines.

LINCON has been extensively tested in an advanced optimal estimation course. The interactive aspects proved highly successful. Hopefully all the "bugs" have been eliminated.

As stated earlier, although the original intent of this thesis was to adapt Desjardins' version of Melsa's LINCON by making it interactive, LINCON began to grow as other routines were added and/or extensively modified.

optcon, the optimal control program using recursive equations derived from dynamic programming, is a new member to the LINCON family. Desjardins' KALMAN, the discrete time Kalman filter program, underwent considerable programming changes before achieving its present form.

#### B. RECOMMENDATIONS

(1) The source program. LINCON FORTRAN A1, must now be passed from user to user and then compiled before it can be used. As can be seen in the terminal session examples, the program was invoked by typing "lincon". Actually, this is an executive program, LINCON EXEC A1, comprising of the following statements:

FILEDEF 09 PRINTER (RECFM FA LRECL 133 BLOCK 133

LOAD LINCON (START

The first statement defines 09 as the printer and permits it to print out 133 characters per line. The second statement invokes the compiled version of LINCON. It is recommended that LINCON be placed on a utility disk so that users may link to it instead of the current procedure.

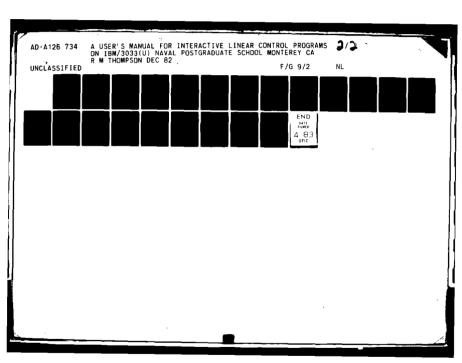
(2) At times it can be exceedingly difficult to interpret the tabular output of some of the programs. It is recommended that a graphics package be developed for RTRESF, RTCATI, KALMAN and OPTCON. The package should be interactive with the output being first displayed on the terminal screen and then allowing the user to choose the type of output, i.e., VERSATEC, TEKTRONIX or print-plot.

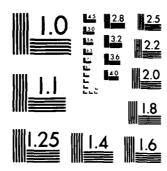
- (3) In general, the programs are limited to eighth-order problems. If the need should arise to solve higher order systems, this limitation may be removed by extending the appropriate dimension statements. The user must remember to alter the format statement pertaining to the output, either decreasing the significant figures or adding a "wrap-around" feature to overcome the printer limitations of 133 characters per line.
- (4) As it is written, LUEN can only solve for a reduced-order observer. It is recommended that the program be modified so that the user has the option of selecting a reduced-order observer or an identity observer.
- (5) The ficticious and real c matrices of STVAR are required to have the Quensions 1  $\times$  n. It is recommended that the program be modified to accept a dimension size of m  $\times$  n.

#### APPENDIX A

```
BASIC MATRIX PROGRAM
PROBLEM IDENTIFICATION:
                                            THESIS EXAMPLE
      0.0
1.000000E+00
1.000000E+00
                             0.0
-2.0000000E+00
-5.000000E-01
       THE DETERMINANT OF THE MAIRIX -1.5 COO OOE+00
      THE INVERSE OF THE MATRIX
1.0000000E+00 -6.66655E-01
0.0 -3.333333E-01
                                                    0.7
6.556665E-01
1.333333E+00
      0.3
0.3
1.330000E+00
      0.0
1.000000E+00
1.00000E+00
      THE MATRIX COEFFICIENT OF S**0
-1.500000E+00
0.0
0.0
5.000000E+00
5.000000E+01
                                                   0.3
-1.300000E+00
-2.330000E+00
   THE EIGENVALJES OF THE A TATRIX
REAL PART IMAGINARY PART
8.228755E-01 0.0
-1.822876E+00 0.0
1.00000E+00 0.0
THE MATRIX COEFFICENT OF EXP( -1.8 0.0 0.0 1.066944 E+00 1.889820E-01
                                          OF EXP( 1.0
3.0
-2.2649775-36
-5.364418E-07
                                            1.0000000E+00) T
0.0
2.861023E-06
07 1.072884E-06
THE MATRIX COEFFICENT

1.000000E+30
```





MICROCOPY RESOLUTION TEST CHART NATIONAL BUREAU OF STANDARDS-1963-A

### APPENDIX B

PARTIAL PRACTION EXPANSION PROBLEM IDENTIFICATION: THESIS EXAMPL	E
THE INPUT FUNCTION JAIN 7.0 COO 00 E+00	
	ERS OF :
NUMERATOR ROOTS ARE REAL PART IMAG. PART 0.0 -1.414213E+00 0.0 1.414213E+00 -1.000000E+00 0.0	
THE INPUT FUNCTION GAIN 7.000000E+00	******
DENOMINATOR COEFFICIENTS - IN ASCENDING P 2.000E+00 1.000E+00 0.0 3.000E+00	OWERS OF S 1.000E+00
DENOMINATOR ROOFS ARE  REAL PART IMAG. PART MULTIPLICE  -3.018859E+00 0.0 1  -9.127604E-01 0.0 1  4.658098E-01 -8.308427E-01 1  4.658098E-01 8.308427E-01 1	TY
<i>·····································</i>	******
RESIDUE MATRIX - REAL PART 5.810872E+00 3.170837E-01 4.360231E-01 4.360231E-01	
RESIDUE MATRIX - IMAG. PART -2.050349E-07 0.0 2.061406E+00 -2.061406E+00	*****

# APPENDIX C

ROOTS OF A POLYN PROBLEM IDENTIFI ***********************************	CATION: THESIS E	ING POWERS OF S
THE ROOTS ARE	REAL PART -1.413221E+00 -1.413221E+00 -1.735564E-01	IMAGINARY PART 1.616248E+30 -1.616248E+33 0.0

#### APPENDIX D

```
RATIONAL TIME RESPONSE PROBLEM IDENTIFICATION:
                                              THESIS EXAMPLE
THE A MATRIX (PLANT MATRIX)
1.0 COO OOE+00 0.0
                                                        0.0
1.000000E+00
1.00000E+00
                            -2.000303E+00
-5.000303E-01
THE B MATRIX (CONTROL VECTOR)
   0.0
1.000000E+00
THE C MATRIX (JUTPUI VECIDR)
1.000000E+00
4.00000E+00
1.00000E+00
THE FEEDBACK COEFFICIENT VECTOR
1.000000E+00
1.000000E+00
THE CONTROLLER GAIN
3.200000E+00
INITIAL CONDITIONS VECTOR-X(0)
0.0
0.0
0.0
THE INPUT FUNCTION GAIN 5.00000E-01
NUMERATOR POLYNOMIAL OF R(S) - ASCENDING POWERS OF S
1.000000E+00 1.00000E+00
NUMERATOR ROOTS ARE REAL PART -1.0000000E+00 0.0
DENOMINATOR POLYNOMIAL OF R(S) - ASCENDING POWERS OF S
-2.000000E+00
DENOMINATOR ROOFS ARE
REAL PART
1.414213E+00 0.0
-1.414213E+00 0.0
THE TIME RESPONSE OF THE STATE X(T)
VECTOR COEFFICIENT OF EXP(-5.00E-01) F*COS(1.20E+00) T
0.0 -3.695463E-01 -9.717697E-01
VECTOR COEFFICIENT OF EXP(-5.00E-01) P*SIN(1.20E+00) T
-3.466749E-01 -7.502007E-02
THE VECTOR CDEFFICIENT OF EXP(
THE VECTOR COEFFICIENT OF EXP(
THE VECTOR CDEFFICIENT OF EXP(
0.0
1.025083E-01
```

THE TIME RESPONSE OF THE DOIPUT Y (T)
THE COEFFICIENT OF EXP (-5.0000E-01) T\*COS (1.2041E+00) T
-1.119588E+00

THE COEFFICIENT OF EXP(-5.3000E-01) T\*SIN(1.2041E+00) T -2.136900E-31

THE COEFFICIENT OF EXP( 1.000000E+00) T 2.6 82209E-06

THE COPPRICIENT OF EXP( 1.414213E+00) T

THE COEFFICIENT OF EXP( -1.414213E+00) T

# APPENDIX E

OBSERVABILITY PROBLEM IDEN	(, CONTROLLABILIT	Y HESIS EXAMPLE
THE A HATRIX 0.0 -1.0000000E+00	(PL ANT MATRIK) 1.000000E+00 -5.000000E-01 0.0	0.0 1.000000E+00 1.000000E+00
2.0000000E+00 0.0 0.0	1.0003 00E+03 1.0003 00E+03	
THE C MATRIX 0.0 1.0000000E+00 -2.000000E+00	1.000000E+03 1.000000E+03 1.000000E+03	0.0 0.0 0.0
OBSERVABILITY THE SYSTEM (	INDEX = 2	LABLE

#### APPENDIX F

```
THESIS ENAMPLE CONTROL/CONTINUOUS KALMAN FILTER PROGRAM THESIS ENAMPLE THESIS ENAMPLE THE STATE OF THE STATE 
     THE B MATRIX (DISTRIBUTION MATRIX)
1.000000E+00 1.000000E+00
     THE Q MATRIX (STATE WEIGHTING MATRIX)
1.000000E+00 1.000000E+00
1.00000E+00 1.000000E+00
                TRANSIENT SOLUTION
TIME = 1.000E+31
GAINS
           TIME = 9.000E+00
GAINS
3.745150E-01 2.
1.404037E-01 1.
                                                                                                      2.808374E-01
1.112729E-01
           TIME = 8.000E+33
GAINS
3.982797E-01 2.
1.424856E-01 1.
                                                                                                      2.849711E-01
1.117260E-01
           TIME 7.09
GAINS
3.99705 dE-01
1.423996E-01
                                                              7.000E+33
                                                                                                      2.847993E-01
1.117371E-01
           TIME = 6.000E+33
GANS
3.997959E-01 2.
1.423879E-01 1.
                                                                                                      2.847759E-01
1.117399E-01
           TIME = 5.000E+00
GAINS
3.997999E-01 2.
1.423877E-01 1.
                                                                                                      2.847756E-01
1.117399E-01
           TIME = 4.000E+33
GAINS
3.997999E-01 2.
1.423877E-01 1.
                                                                                                      2.847756E-01
1.117399E-01
```

```
TIME = 3.000E+00
    GAINS
3.997999E-01
1.423877E-01
                                        2.847756E-01
1.117399E-01
     TIME = 2.000E+00
GAINS
3.997999E-01 2.
1.423877E-01 1.
                                        2.847756E-01
1.117399E-01
     TIME = 1.000E+33
GAINS
3.997999E-01 2.
1.423877E-01 1.
                                        2.847756E-01
1.117399E-01
   TIME = 2.861E-J5
GAINS
3.997999E-01 2.847756E-01
1.423877E-01 1.117399E-01
       *************
       *** NCITER NETTIT ***
  THE P MATRIX (INITIAL BOUNDARY VALUE MATRIX)
0.0
0.0
0.0
0.0
0.0
0.0
      TRANSIENT SOLUTION
TIME = 0.0
GAINS
0.0
                      0.0
    TIME = 5.000E-31
GAINS
2.393162E-01 2.
8.018157E-01 5.
                                        2.004539E-01
5.849594E-01
    TIME = 1.000E+00
GAINS
2.708317E-01 2.
8.511767E-01 7.
                                        2. 127941E-01
7. 053521E-01
     TIME = 1.500E+00
GAINS
2.768488E-01
8.532724E-01
7.
                                        2. 133181E-01
7. 055743E-01
     TIME = 2.000E+03
GAINS
2.784715E-01 2.
8.530354E-01 7.
                                        2. 132589E-01
7. 055843E-01
     TIME = 2.500E+33
GAINS
2.789688E-01 2.
8.528843E-01 7.
                                        2. 132211E-01
7. 055938E-01
     TIME = 3.000E+33
GAINS
2.791271E-01
8.528297E-01
7.
                                    2.132074E-01
7.055972E-01
     TIME = 3.500E+00
GAINS
2.791781E-01 2.
8.528109E-01 7.
                                        2. 132027E-01
7. 055984E-01
```

TIME = 4.000E+33
2.791938E-01 2.132320E-01
8.528078E-01 7.055986E-01
TIME = 4.500E+33
2.791983E-01 2.132018E-01
8.528070E-01 7.055986E-01
TIME = 5.000E+33
2.791983E-01 2.132018E-01
8.528070E-01 7.055986E-01
8.528070E-01 7.055986E-01
8.528070E-01 7.055986E-01

## APPENDIX G

```
THE GAMMA MATRIX (DISTRIBUTION MATRIX)
1.2500000-01
5.0000000-01
 THE W MATRIX (MEAN-SQUARE MAGNITUDE OF THE PERTURBATION ACCELERATION MATRIX)
 THE R MATRIX (MEASUREMENT NOISE COVARIANCE MATRIX) 5.0000000+00
THE P MATRIX (INITIAL CONDITION MATRIX)
1.000000E+03
0.0
1.000000E+03
 K = 0
GAINS
9.950248E-01
                     0.0
                     1.923759E+03
 K = 2
GAINS
8.290299E-01
                     9.980751E-01
 K = 3
GAINS
7.028207E-01
                     5. 237227E-01
 GAINS
6.134719E-01
                     4.522047E-01
 K = 5
GAINS
5.543699E-01
                     3.716016E-01
 K = 6
GAINS
5.184991E-01
                     3. 362513E-01
 K = 7
GAINS
4.992278E-01
                     3.232893E-01
 K = 8
GAINS
4.90446 8E-01
                     3. 201340E-31
 K = 9
GAINS
4.872549E-01
                     3.202326E-01
```

K = 10 GAINS	
4. 8644 0 0E-01	3.207918E-01
K = 11 GAINS	
GAINS 4.863475E-01	3. 2103 92E-01
K = 12 GAINS	
4. 863456E-01	3. 2099 99E-01
K = 13 GAINS 4.862944E-01	3 2005 (35 04
K = 14	3. 208563E-01
GÂINS 4.862168E-01	3. 2072 75E-01
K = 15	
GAINS 4.861480E-01	3. 2064 94E-01
K = 16	
GAINS 4.861035E-01	3.206143E-01
K = 17 GAINS	
GAINS 4.860806E-01	3. 206035E-01
K = 18 GAINS	3 00 (00 ) = 04
4.860712E-01	3. 206024E-01
K = 19 GAINS 4.860685E-01	3. 206034E-01
K = 20	3. 20003-11-01
GAINS 4.860680E-01	3.206041E-01
* * * * * * * * * * * * * * * *	**** **** ****

#### APPENDIX H

```
STATE VARIABLE FEEDBACK PROGRAM PROBLEM IDENTIFICATION: THE
                                             THESIS EXAMPLE
  0.3
1.000000E+00
  THE B MATRIX (CONTROL MATRIX)
     0.0
1.000000E+00
  OPEN-LOOP CALCULATIONS
DENOMINATOR CORPFICIENTS - IN ASCENDING POWERS OF S
3.0000E+00 3.0000E+00 1.0000E+00
                                                             IMAGINARY PART
-1.732051E+00
1.732051E+00
                                     REAL PART
  THE ROOTS ARE
                                    -1.500000E+00
  THE C MATRIX (PICTICIOUS DUIPUT VECIOR)
0.0
1.000000E+00
  NUMERATOR COEFFICIENTS - IN ASCENDING POWERS OF S 0.0 1.000000E+00
                                REAL PART
-1.300000E+00
3.3
                                                               IMAGINARY PA.
0-0
0-0
0-0
  THE ROOTS ARE
  THE C MATRIX (DUTPUT VECTOR)
1.000000E+00
1.000000E+00
  NUMERATOR COEFFICIENTS - IN ASCENDING POWERS OF S 2.000000E+00
                                                               IMAGINARY PART
0.0
                                    REAL PART
  THE ROOTS ARE
  CLOSED-LOOP CALCULATIONS
KEY = P *****
THE NUMERATOR OF H-EQUIVALENT -
IN ASCENDING POWERS OF S
5.000000E-01 1.500000E+00
                                                     1.500000E+00
                                 REAL PART INAG. PART -5.303000E-01 -2.886756E-01 2.886756E-01
   THE ROOTS ARE
  THE FEEDBACK CDEFFICIENTS
5.000000E-01
                                                       1.500000E+00
   THE GAIN =
                        2.000000E+00
   THE CLOSED-LOOP CHARACTERISTIC POLYNOMIAL -
IN ASCENDING POWERS OF S
4.0000E+00 6.0000E+00 4.0000E+00
                                                4.0000E+00 1.0000E+00
```

THE ROOTS ARE

REAL PART IMAG. PART
-2.300000E+00 0.0
-1.300000E+00 -1.000000E+00

MAXIMUM NORMALIZED ERROR = 0.0

#### APPENDIX I

```
STATE VARIABLE FEEDBACK P3 ) GRAM
PROBLEM IDENTIFICATION: THESIS EXAM
THE A MATRIX (PLANT MATRIX)
0.0 1.000000E+00 0.0
0.0 0.0 1.000000E+00
0.0 0.0 0.0 0.0
0.0 -1.500000E+01 -2.300000E+01
                                                            THESIS EXAMPLE
                                                                                       0.0
0.0
1.000000 E+00
-9.000000 E+00
     THE B MATRIX (CONTROL MATRIX)
0.0
0.0
0.0
1.000000E+00
     OPEN-LOOP CALCULATIONS
DENOMINATOR COEFFICIENTS - IN ASCENDING POWERS OF S
0.0 1.500E+31 2.300E+01 9.000E+00 1.000E+00
                                                 REAL PART
-3.300000E+00
-4.999999E+00
-9.999999E-01
      THE ROOTS ARE
                                                                                   IMAGINARY PART
0.0
0.0
                                                                                   0.0
     THE C NATEIX (JUTPUT VESTOR)
2.000300E+31
1.000300E+31
0-0
     NUMERATOR COEFFICIENTS - IN ASCENDING POWERS OF S 2.000000E+31 1.000005E+01
      THE ROOTS APE
                                                 -2.00000E+00
                                                                                   IMAGINARY PART
     CLOSED-LOOP CALCULATIONS

KEY = P *****

THE NUMERATOR OF H-EQUIVALENT-IN ASCENDING POWERS OF S

2.000E+01 1.300E+01 -6.000E+00 -3.000E+00
                                             REAL PART IMAG. PART
-1.039623E+00
3.206289E+00 0.0
      THE ROOTS ARE
     THE FEEDBACK COEFFICIENTS 2.000E+01 1.300E+01 -6.000E+00 -3.000E+00
                                 1.000000E+0)
      THE GAIN =
      THE CLOSED-LOOP CHARACTERISTIC POLYNOMIAL IN ASSENDING POWERS OF S 0.300E+01 1.700E+01 6.300E+00
2.000 E+0 1
                                                                                               1.000E+00
                                             REAL PART IMAG. PART
-1.000000E+00 -2.00000E+00
-2.00000E+00 0.0
-1.399516E+00 0.0
      THE ROOTS ARE
      MAXIMUM NORMALIZED ERROR =
     OBSER VABILITY, CONTROLLABILITY PROBLEM IDENTIFICATION: THESIS EXAMPLE
```

```
THE A MATRIX (PLANT MATRIX)
                                                                                                                                                                                                                                       0.0
0.0
1.000000 E+00
-9.000000 E+00
000
                                                                                                                                                1.000000E+00

-2.300000E+01
                                                           0.0
-1.5000 00 E+01
THE B MATRIX
0.0
0.0
0.0
1.00000000+00
 1.0000 CE+00
                                                                                     0. 0
1. 000000E+00
 0.0
          OBSERVABILITY IN DEX = 3
THE SYSTEM (A, B) IS CONTROLLABLE
         LUENBERGER OBSERVER DESIGN PROGRAM
PROBLEM IDENTIFICATION: THESIS EXAMPLE
           THE A MATRIX (PLANT MATRIX)
0.0 1.005000E+00 5
0.0 0.0 1
0.0 0.0 2
0.0 0.0 2
0.0 0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
0.0 2
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0.0 2
0.0 2
0.0 2
0.0 2
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0.0 2
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1.0000000 E+00
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                                                                                                                                                 -2.300000E+01
           THE B MATRIX (DISTRIBUTION MATRIX) 0.0
                      0.0
1.0 COO OOE+00
            THE C MATRIX (JUTPUT MAIRIX)
1.0 COO OOE+00
1.0 00000E+00
            THE DESIRED FEEDBACK COEFFICIENTS
2.000000E+01
1.300000E+01
-6.00000E+00
-3.00000E+00
            OBSERVER EIGENVALUES
                                                                                                                                                                  REAL PART
-3.500000E+00
-4.000000E+00
                                                                                                                                                                                                                                                                                         PART
            OBSERVER CHARACTERISTIC POLYNOMIAL COEFFICIENTS IN ASCENDING POWERS OF S 1.4 COOOCE+01 7.500000 E+00 1.00000 E+00
            THE F MATRIX (JBSERVER EIGENVALUE MAIRIX)
-7.5 COO 00E+00 1.000 JOJE+00
-1.4 000 00E+31 0.0
             THE G 1 NATRIX (DBSERVER GAIN MATRIX)
8.549997E+01 2.924994E+01
             THE G2 MATRIX (DBSERVER GAIN MATRIX)
-2.999997E+00
-1.500001E+J0
```

OUTPUT FEEDBACK COEFFICIENTS 2.000000E+01 8.500004E+00 COMPRISATOR FEEDBACK COEFFICIENTS 1.000000E+00 0.0

## APPENDIX J

```
OPTIMAL CONTROL PROGRAM
PROBLEM IDENTIFICATION:
                                                                 THESIS EXAMPLE
  THE NUMBER OF TIME INTERVALS =
  THE ORDER OF THE SYSTEM = 2
  THE 0 MATRIX
0.0 COOOE+00
                                          0.0
2.000000E+00
  THE SCALAR R =
                                        1.
  THE SAMPLE INTERVAL (DT) =
                                                                       0.1000
  THE A MATRIX (PLANT MATRIX)
1.000000E+00 0.0
  THE B MATRIX (DISTRIBUTION MATRIX)
0.0
1.000000E+00
  THE PHI MATRIX
1.005003E+00
1.001667E-01
                                          1.001667E-01
1.005003E+00
  THE DEL MATRIX
5.004164E-03
1.001667E-01
MINIMIZATION OVER ALL STAGES
U(1) = AT(J) * X(J)
AT(1) = -2.46 E - 3.2
AT(2) = -1.98 E - 3.1
U(2) = AT(J) *X(J)
AT(1) = -7.72E-32
AT(2) = -3.91E-31
U(3) = AT(J) *X(J)
AT(1) = -1.55 E-0.1
AT(2) = -5.79 E-0.1
U(4) = AT(J) *X(J)
AT(1) = -2.54 = -0.1
AT(2) = -7.58 = -0.1
U(5) = AT(J) *X(J)
AT(1) = -3.70E-01
AT(2) = -9.29E-01
U(6) = AT(J) * (J)
AT(1) = -4.98E-J1
AT(2) = -1.09E+J0
U(7) = AT(J) *X(J)
AT(1) = -6.34 E-31
AT(2) = -1.24 E+30

    \begin{array}{rcl}
      0 & (8) & = & \text{AT}(J) *X(J) \\
      \text{AT}(1) & = & -7.74 = -31 \\
      \text{AT}(2) & = & -1.38 = +30 \\
    \end{array}
```

U(9) = AT(J) \*X(J) AT(1) = -9.12E-01 AT(2) = -1.50E+00U(10) = AT(J) \*X(J) AT(1) = -1.05E+00 AT(2) = -1.61E+00U(11) = AT(J) \*X(J) AT(1) = -1.18E+00 AT(2) = -1.72E+00U(12) = AT(J) \*X(J) AT(1) = -1.30E+30 AT(2) = -1.81E+00U(13) = AT(J) \*X(J) AT(1) = -1.41E+00 AT(2) = -1.89E+00U(14) = AT(J) \*X(J) AT(1) = -1.51E+00 AT(2) = -1.96E+00U(15) = AT(J) \*X(J) AT(1) = -1.60E+10 AT(2) = -2.02E+30U(16) = AT(J) \*X(J) AT(1) = -1.68E+30 AT(2) = -2.07E+30U(17) = AT(J) \*X(J) AT(1) = -1.75E+30 AT(2) = -2.12E+30U(18) = AT(J) \*X(J) AT(1) = -1.822+30 AT(2) = -2.16E+30U(19) = AT(J) \*X(J) AT(1) = -1.88E+00 AT(2) = -2.20E+00U(20) = AT(J) \*X(J) AT(1) = -1.93E+30 AT(2) = -2.23E+30U(21) = AT(J) \*X(J) AT(1) = -1.97E+30 AT(2) = -2.25E+30U(22) = AT(J) \*X(J) AT(1) = -2.01E+30 AT(2) = -2.27E+30U(23) = AT(J) \*X(J) AT(1) = -2.04 E+30 AT(2) = -2.29 E+30U(24) = AT(J) \* K(J)
 AT(1) = -2.07E+00
 AT(2) = -2.31E+00U(25) = AT(J) \*X(J) AT(1) = -2.09E+30 AT(2) = -2.32E+30

```
U(26) = AT(J) + X(J)
AT(1) = -2.11E+10
AT(2) = -2.34E+00
U(27) = AT(J) *X(J)
AT(1) = -2.13E+30
AT(2) = -2.35E+30
U(28) = AT(J) *X(J)
AT(1) = -2.15E+30
AT(2) = -2.35E+30
U(29) = AT(J) + X(J)
AT(1) = -2.16E+30
AT(2) = -2.36E+30
U(30) = AT(J) *X(J)
AT(1) = -2.17E+30
AT(2) = -2.37E+30
U(31) = AT(J) *K(J)
AT(1) = -2.18E+00
AT(2) = -2.37E+00
U(32) = AT(J) *X(J)
AT(1) = -2.19E+00
AT(2) = -2.38E+00
U(33) = AT(J) *X(J)
AT(1) = -2.20 E+J0
AT(2) = -2.38 E+J0
U(34) = AT(J) *X(J)
AT(1) = -2.20E+30
AT(2) = -2.39E+00
U(35) = AT(J) *X(J)
AT(1) = -2.21E+30
AT(2) = -2.39E+30
U(36) = AT(J) *X(J)
AT(1) = -2.21E+30
AT(2) = -2.39E+30
\begin{array}{lll} U & (37) & = & \text{AT} (J) * (37) \\ \text{AT} & (1) & = & -2.22 \text{ E} + 30 \\ \text{AT} & (2) & = & -2.39 \text{ E} + 30 \end{array}
\begin{array}{lll} U & (38) & = & \text{AT}(J) * X & (J) \\ \text{AT}(1) & = & -2.22 \pm +30 \\ \text{AT}(2) & = & -2.43 \pm +30 \end{array}
U(39) = AT(J) *X(J)
AT(1) = -2.22 E+30
AT(2) = -2.40 E+30
U(49) = AT(J) *X(J)
AT(1) = -2.23E+30
AT(2) = -2.40E+30
```

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